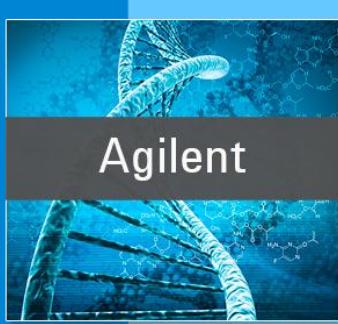


Métodos de Caracterización Dieléctrica



Adolfo Del Solar

Application Engineer
adolfo_del-solar@agilent.com

Agenda



Unlocking Measurement Insights for 75 Years

- **Introduction**
- **Permittivity and permeability**
- **Measurement techniques and systems**
- **Parallel plate**
- **Coaxial probe**
- **Transmission line**
- **Free space**
- **Resonant cavity**

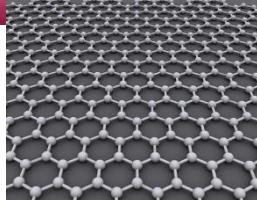
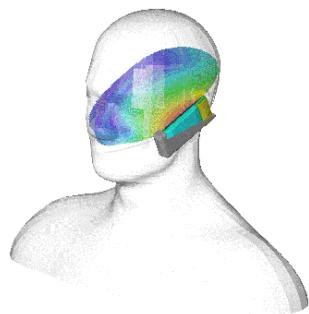
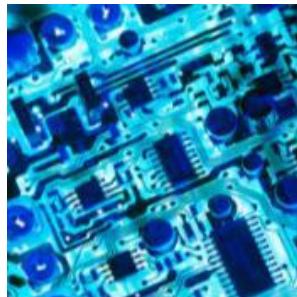
Agenda



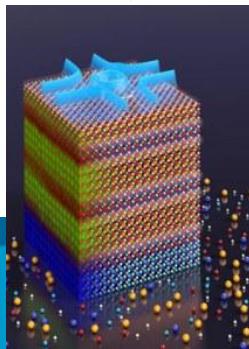
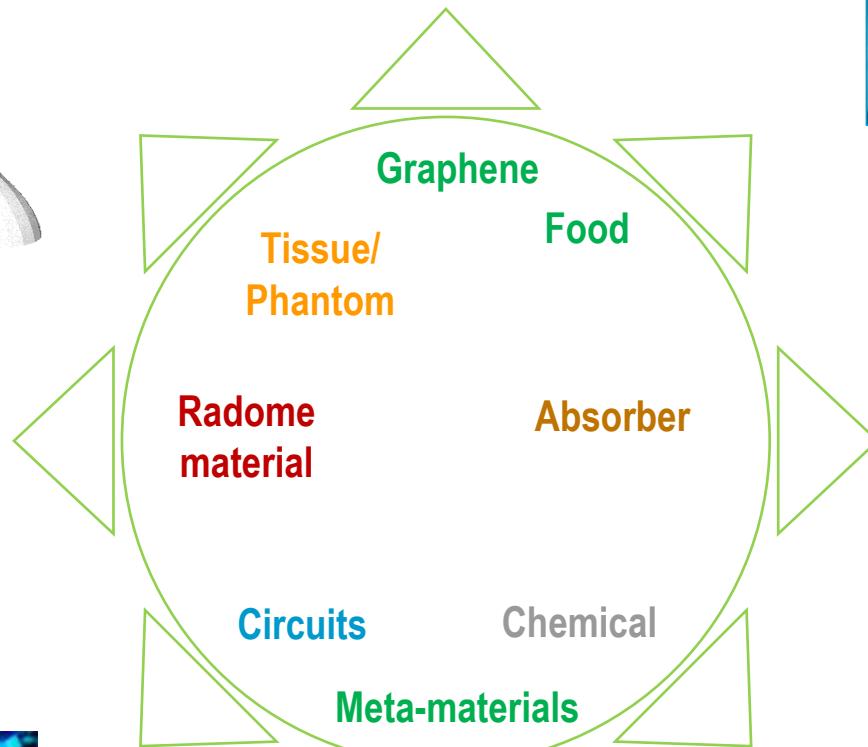
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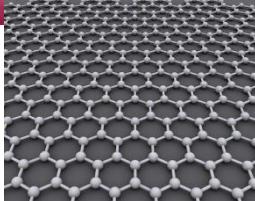
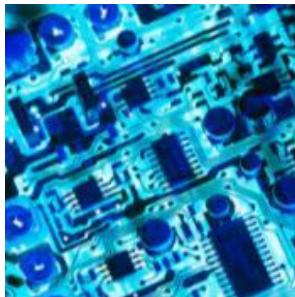
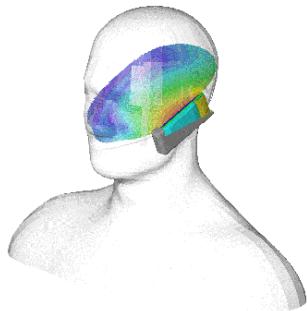
Type of material



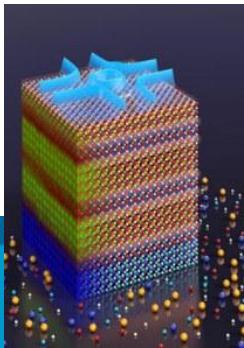
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Type of material



Industry	Application/Products
Electronics	Capacitors, substrates, PCB antennas, ferrites, absorbers, SAR phantom materials
Aerospace/ Defense	Stealth, RAM (radiation absorbing materials), radomes
Industrial/ Materials	Ceramics & composites: A/D and automotive components, coatings Polymers & plastics: Fibers, films, Insulation materials Hydrogel: Disposable diaper, soft contact lens Liquid crystal: Displays Other products containing these materials: Tires, paint, adhesives, etc.
Food & Agriculture	Food preservation (spoilage) research, food development for microwave, packaging, moisture measurements
Mining	Moisture measurements in wood or paper, oil content analysis
Pharmaceutical & Medical	Drug research and manufacturing, bio-implants, human tissue characterization, biomass, fermentation



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Permittivity (Dielectric Constant)

$$\kappa = \frac{\epsilon}{\epsilon_0} = \epsilon_r = \epsilon_r' - j\epsilon_r''$$

interaction of a material in the presence
of an external electric field.

Permittivity (Dielectric Constant)

$$\kappa = \frac{\epsilon}{\epsilon_0} = \epsilon_r = \epsilon_r' - j\epsilon_r''$$

interaction of a material in the presence of an external electric field.

Permeability

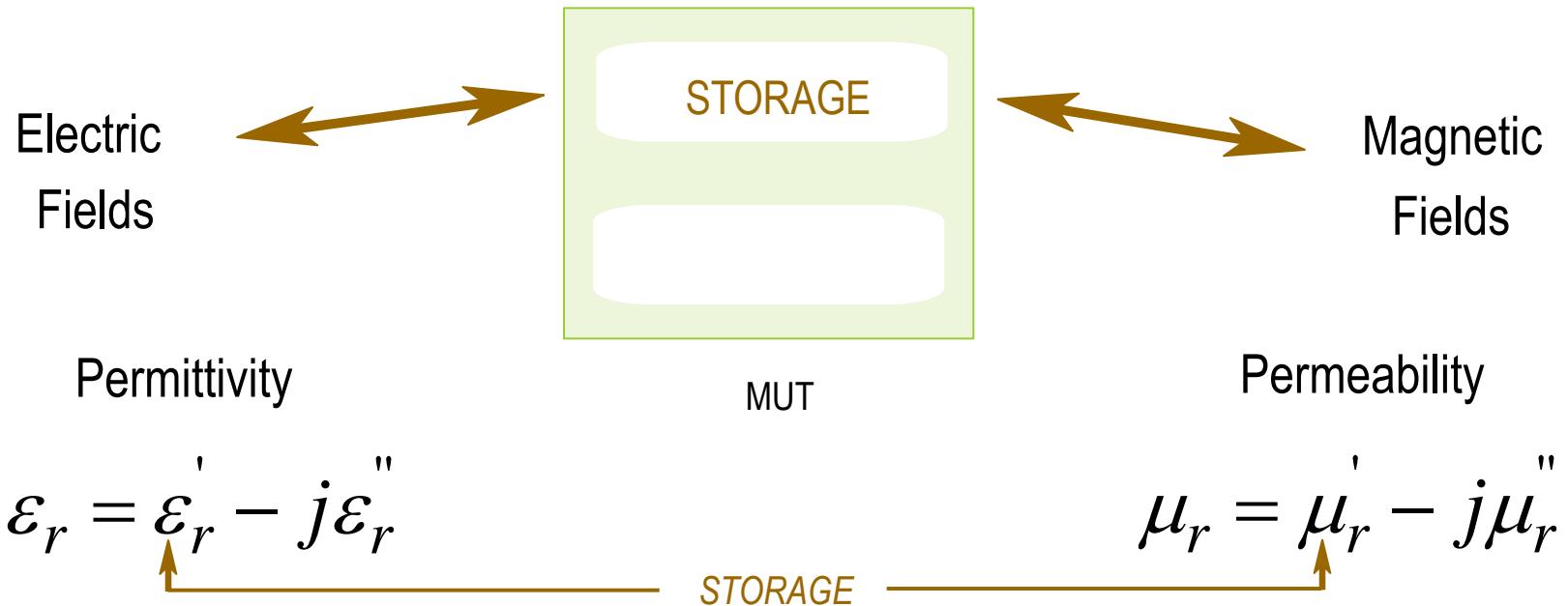
$$\mu = \frac{\mu}{\mu_0} = \mu_r = \mu_r' - j\mu_r''$$

interaction of a material in the presence of an external magnetic field.

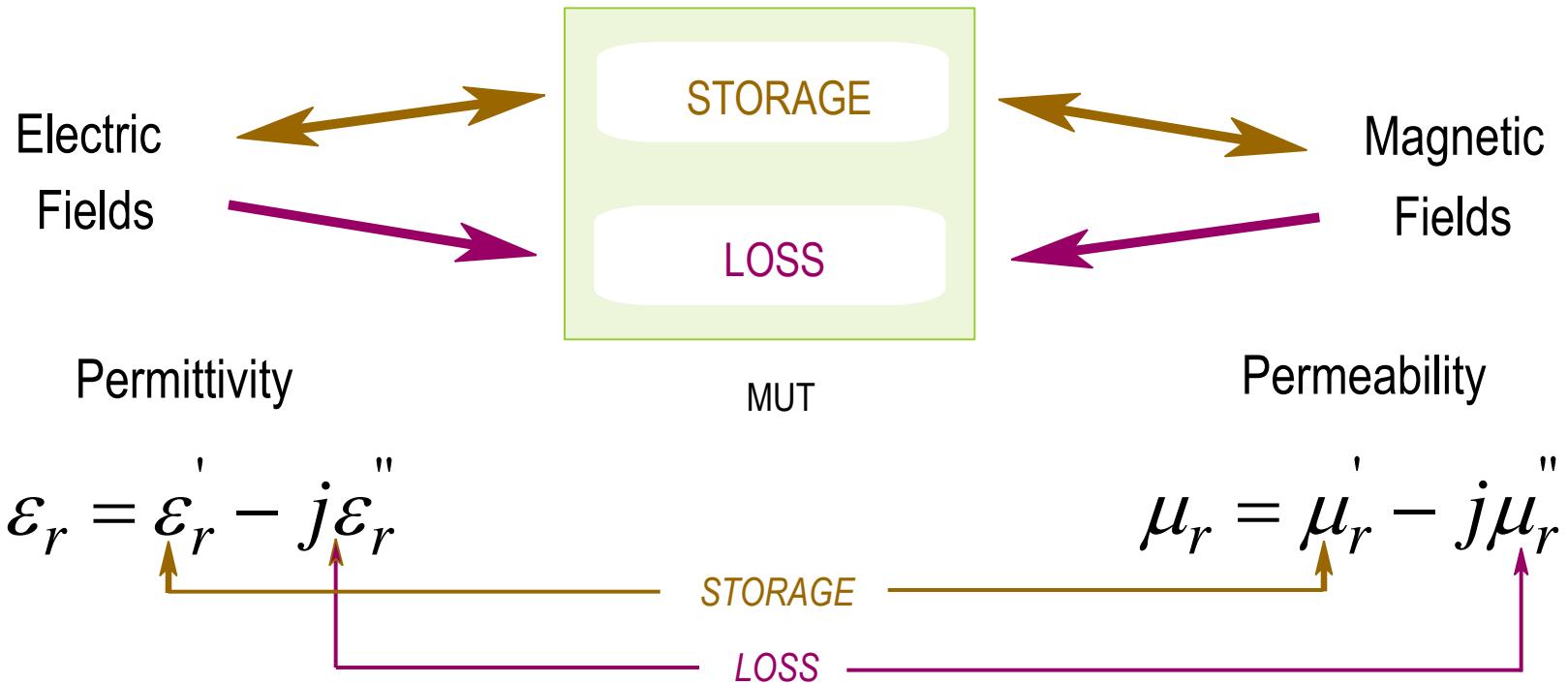
Electromagnetic field interaction



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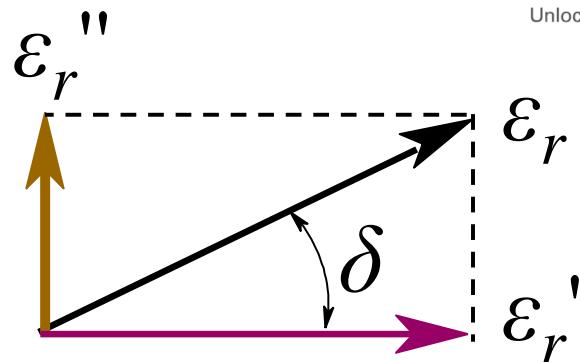


Electromagnetic field interaction



Loss tangent

$$\tan \delta = \frac{\epsilon_r''}{\epsilon_r'}$$



$$\tan \delta = D = \frac{1}{Q} = \frac{\text{Energy Lost per Cycle}}{\text{Energy Stored per Cycle}}$$

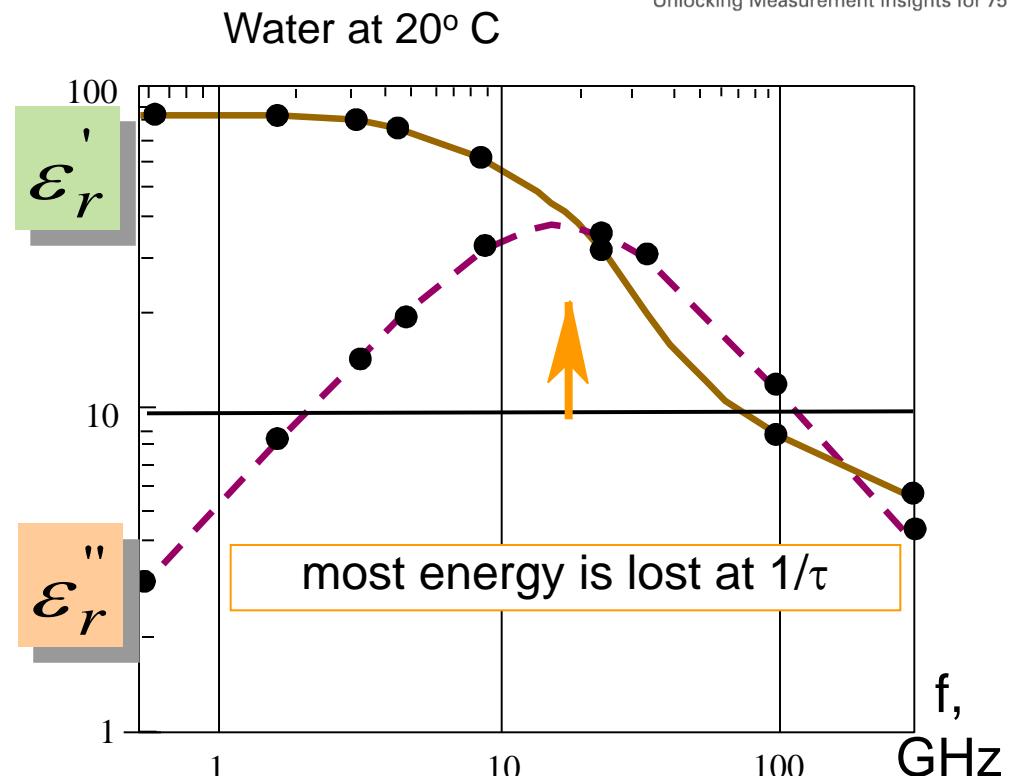
 **D** Dissipation Factor  **Q** Quality Factor

 **Df**

Relaxation constant

τ = Time required for $1/e$ of an aligned system to return to equilibrium or random state, in seconds.

$$\tau = \frac{1}{\omega_c} = \frac{1}{2\pi f_c}$$



Debye equation : $\epsilon(\omega) = \epsilon_{\infty} + \frac{\epsilon_s - \epsilon_{\infty}}{1 + j\omega\tau}$

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Material evaluation measurement system



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Composed by 3 main pieces:

- Precise measurement instruments
- Test fixtures that hold the MUT
- Software that can calculate & display basic material parameters

The measurement instrument and the test fixtures are determined by the **measurement technique** chosen.

Measurement techniques

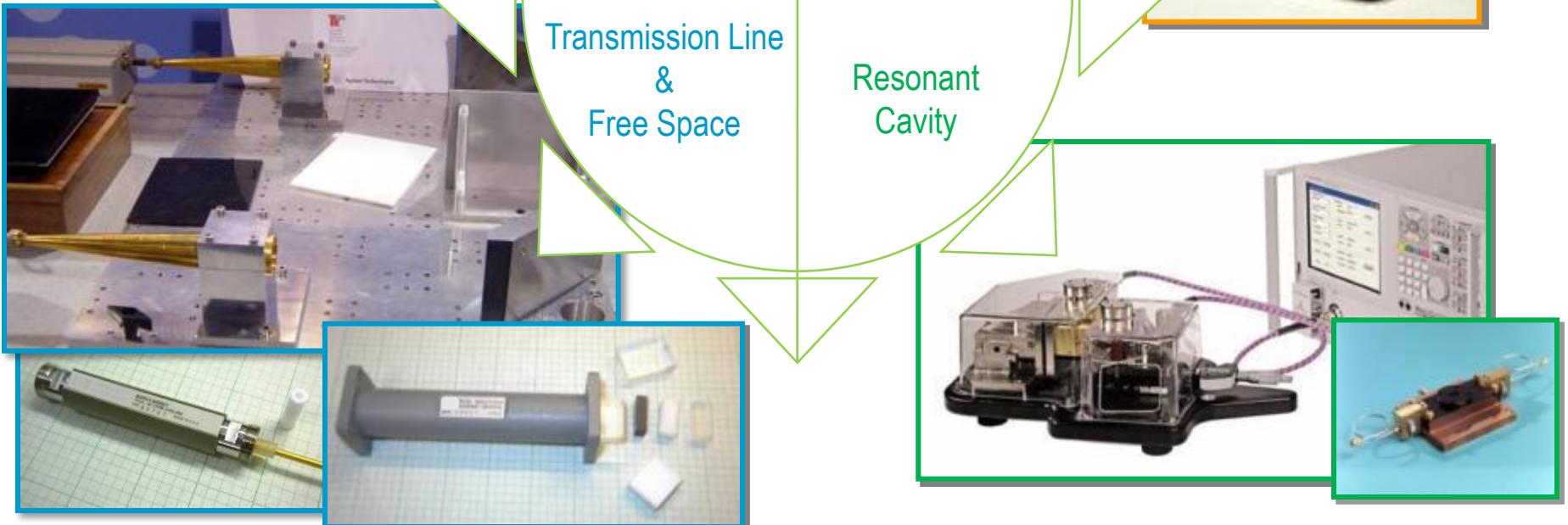


Parallel
Plate

Coaxial
Probe

Transmission Line
&
Free Space

Resonant
Cavity



Which method is best?



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It Depends...



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Which method is best?



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It Depends... on

- ✓ Frequency of interest
- ✓ Expected value of ε_r and μ_r
- ✓ Required measurement accuracy

Which method is best?



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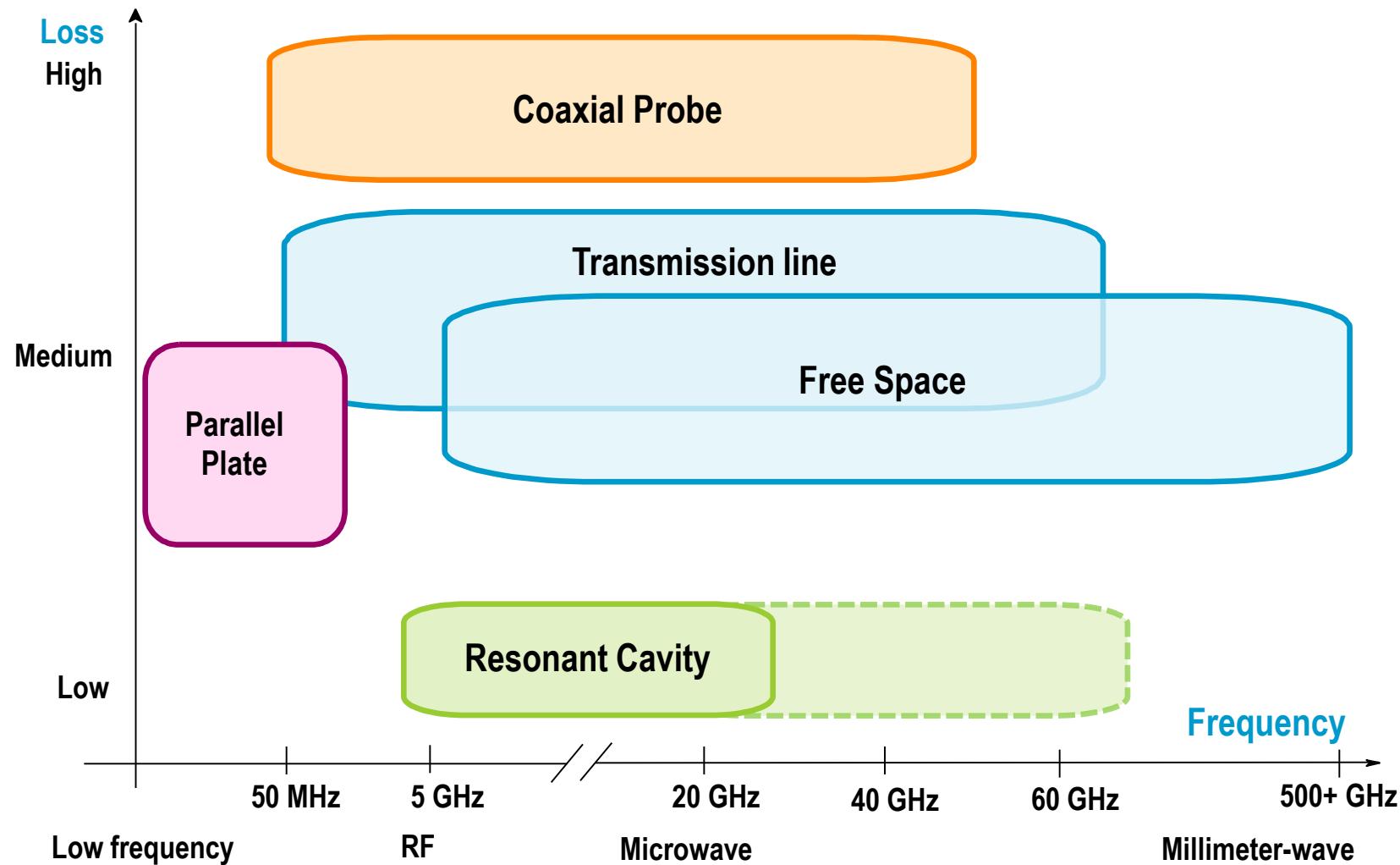
It Depends... on

- ✓ Frequency of interest
- ✓ Expected value of ϵ_r and μ_r
- ✓ Required measurement accuracy
- ✓ Material properties (i.e., homogeneous, isotropic)
- ✓ Form of material (i.e., liquid, powder, solid, sheet)
- ✓ Sample size restrictions

It Depends... on

- ✓ Frequency of interest
- ✓ Expected value of ϵ_r and μ_r
- ✓ Required measurement accuracy
- ✓ Material properties (i.e., homogeneous, isotropic)
- ✓ Form of material (i.e., liquid, powder, solid, sheet)
- ✓ Sample size restrictions
- ✓ Destructive or non-destructive
- ✓ Contacting or non-contacting
- ✓ Temperature

Measurement methods vs. frequency and material loss

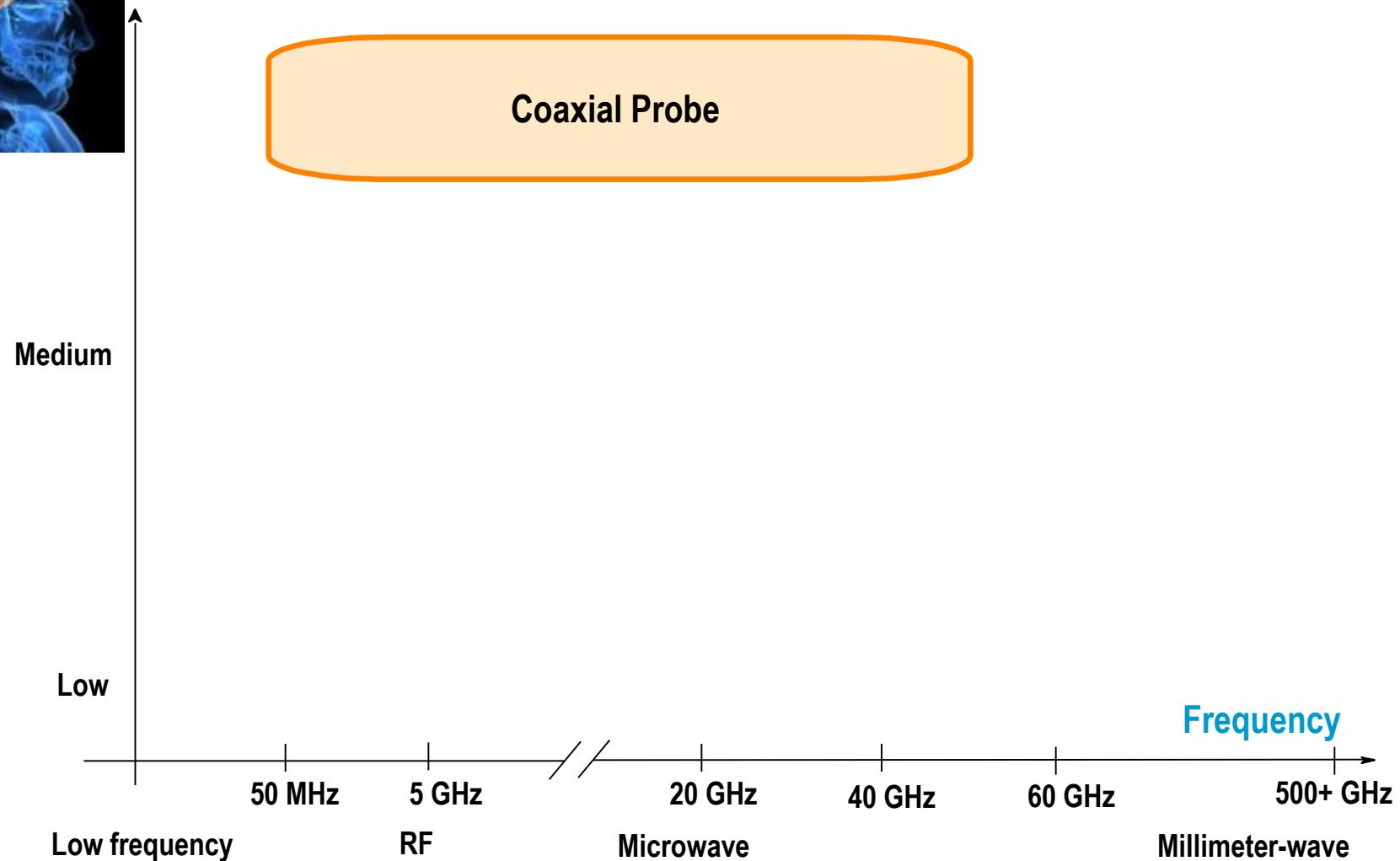




Measurement methods vs. frequency and material loss



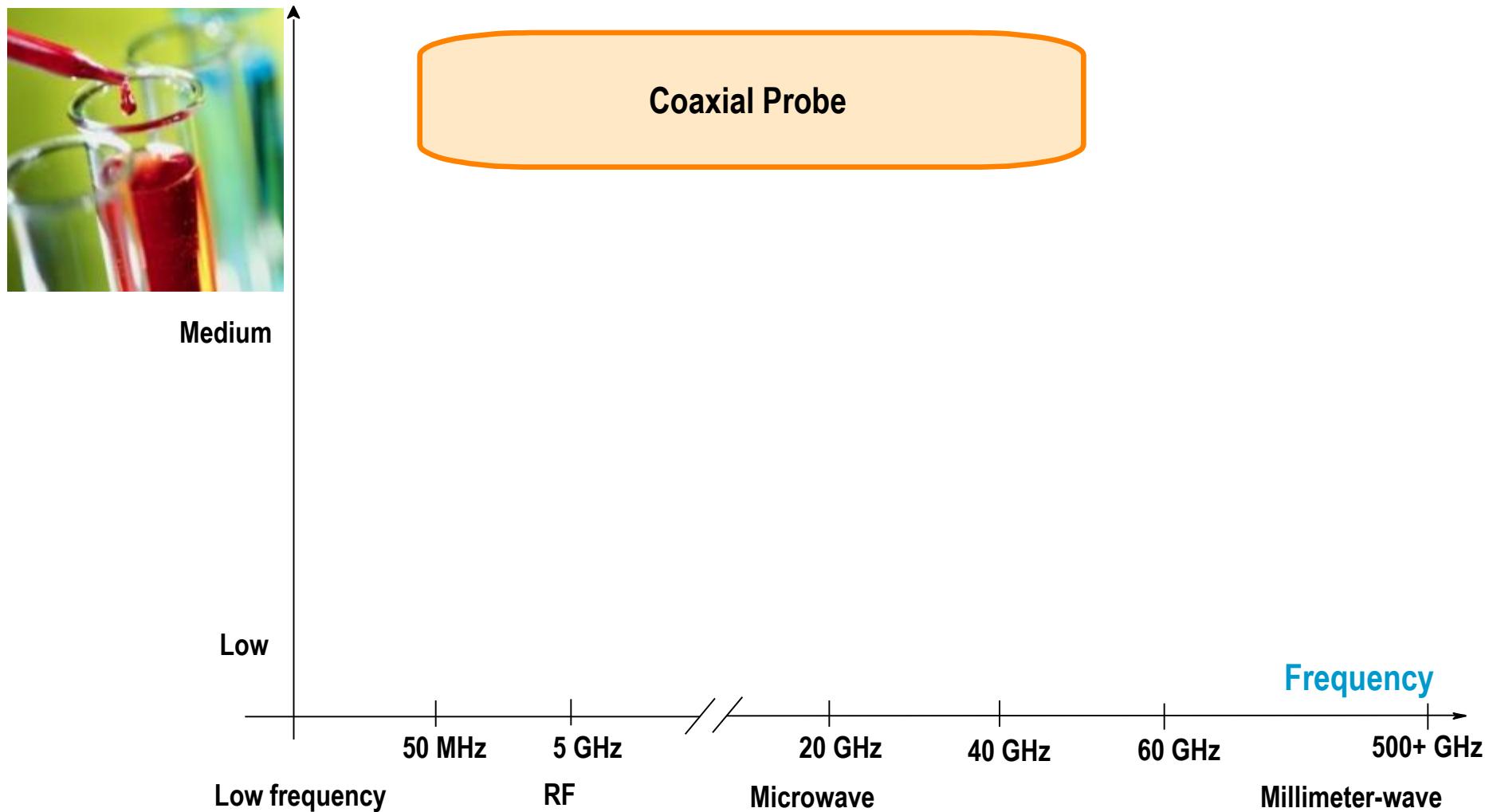
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Measurement methods vs. frequency and material loss



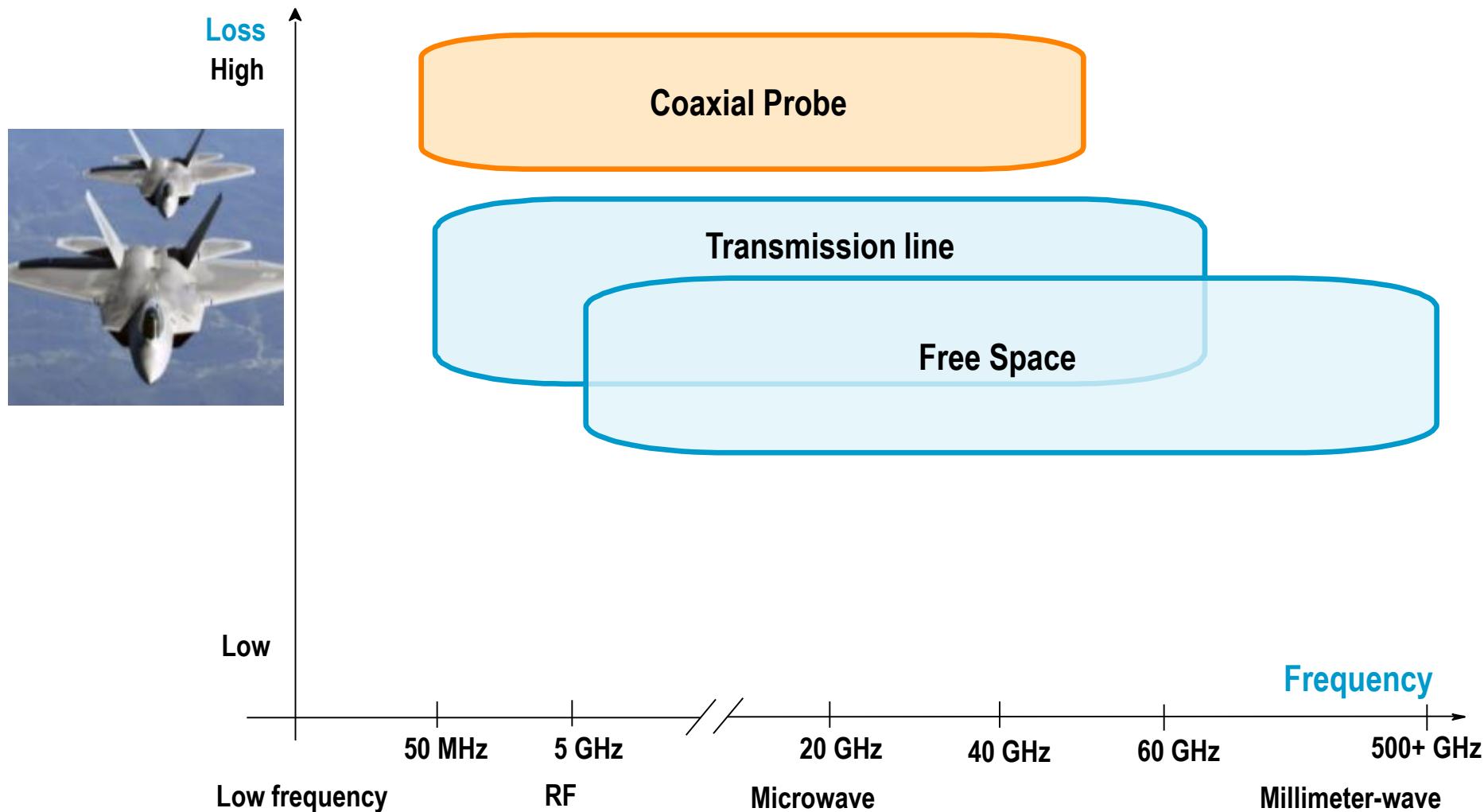
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Measurement methods vs. frequency and material loss



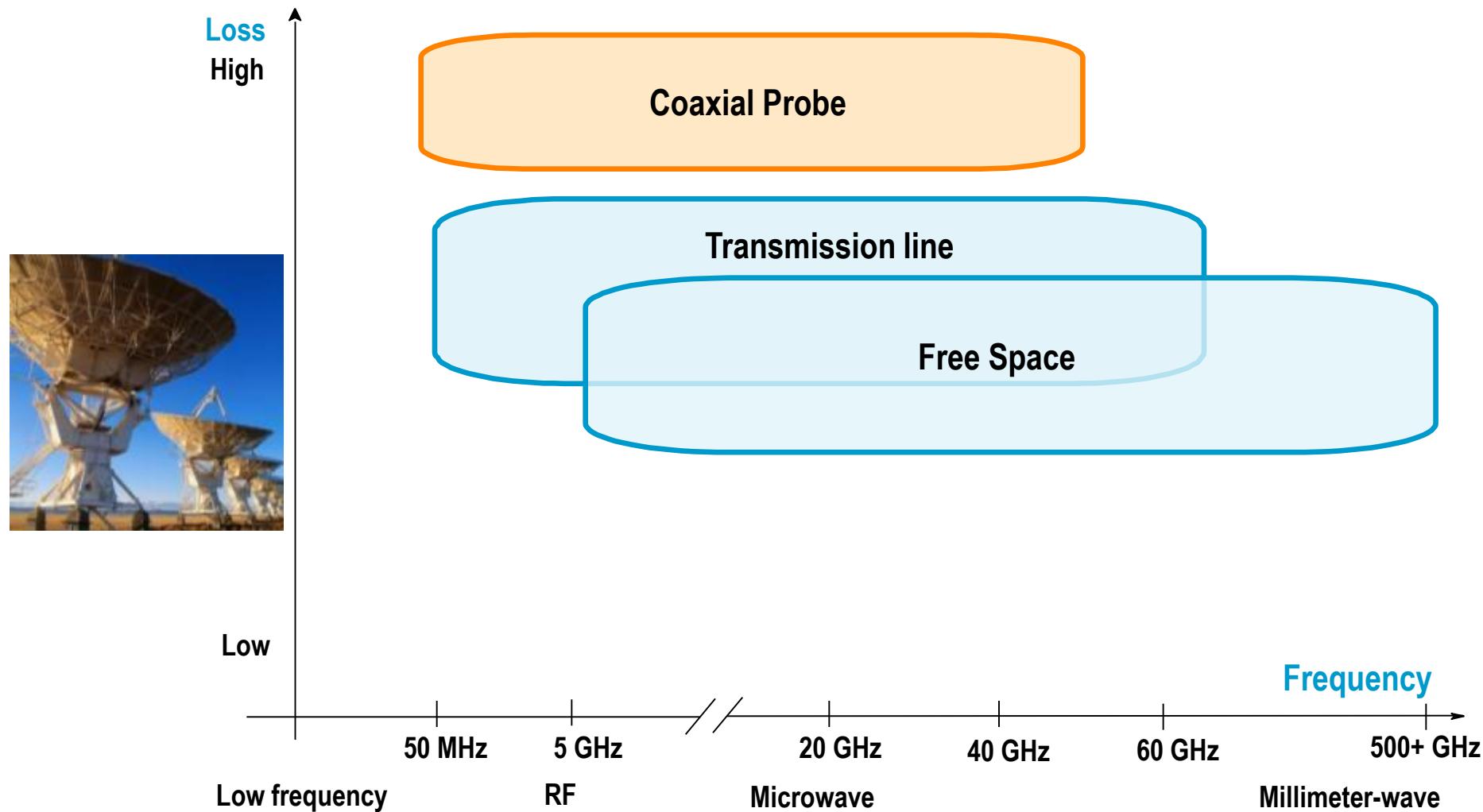
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Measurement methods vs. frequency and material loss



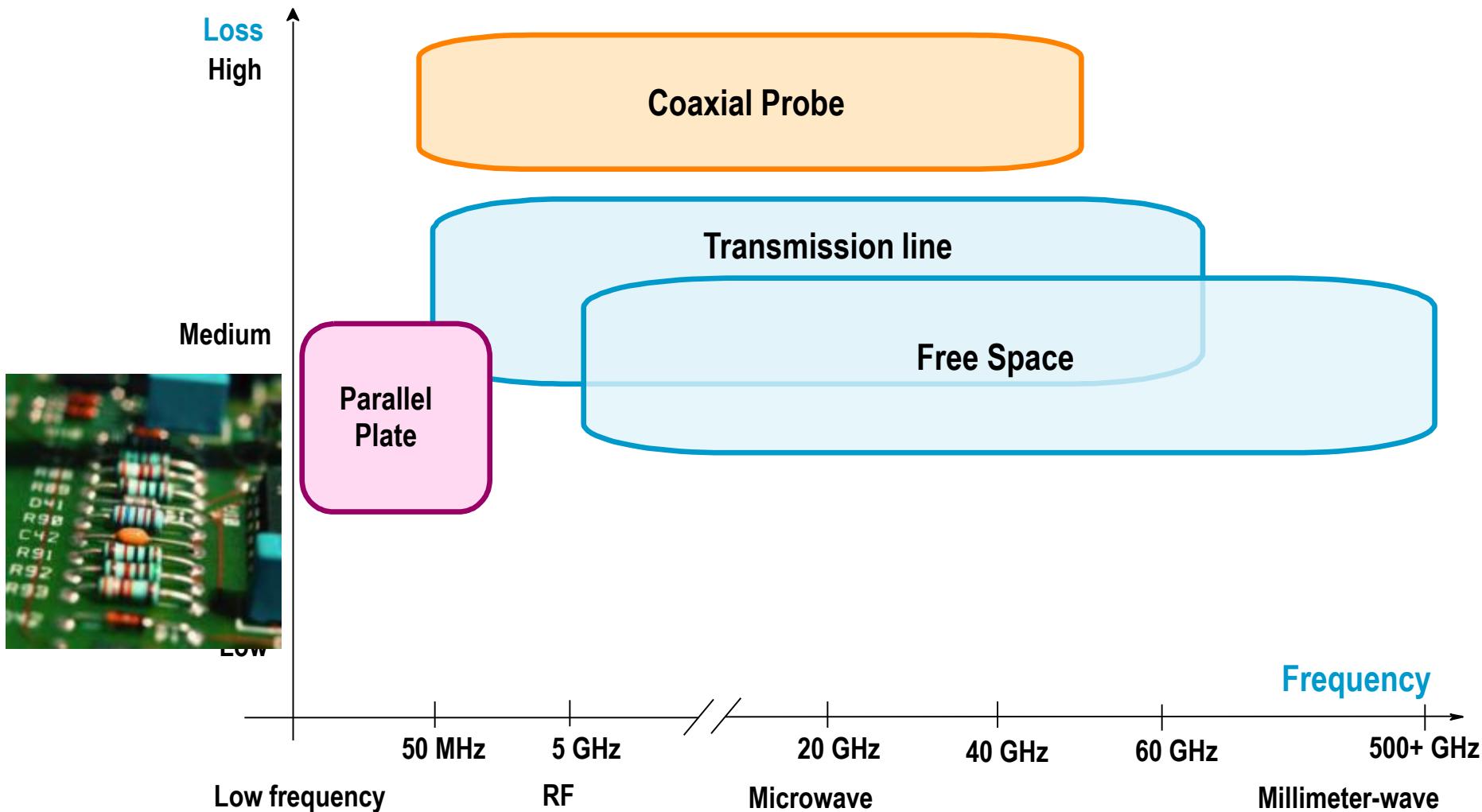
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Measurement methods vs. frequency and material loss



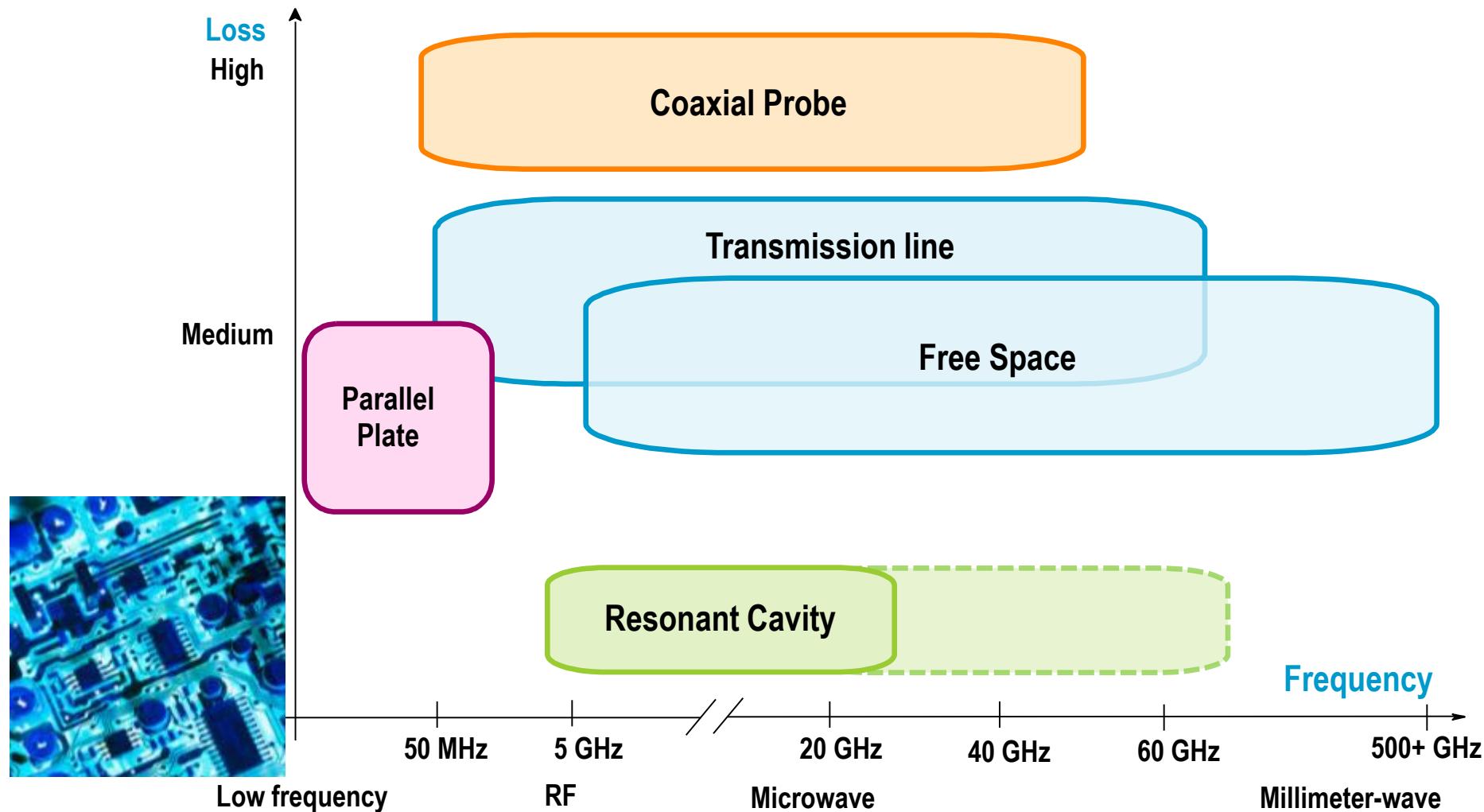
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Measurement methods vs. frequency and material loss



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Agenda



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Parallel plate capacitor system



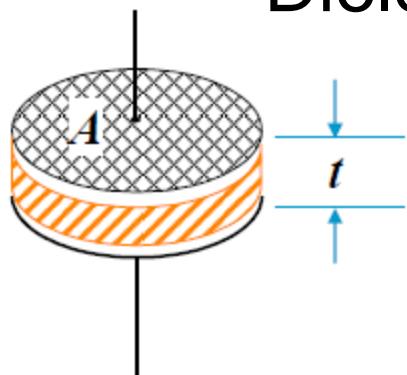
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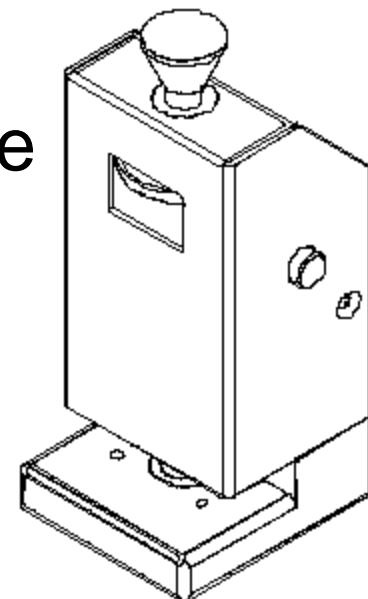
$$\epsilon'_r = \frac{C}{\epsilon_0 \frac{A}{t}}$$

$$\tan \delta = D$$

Dielectric Test Fixture

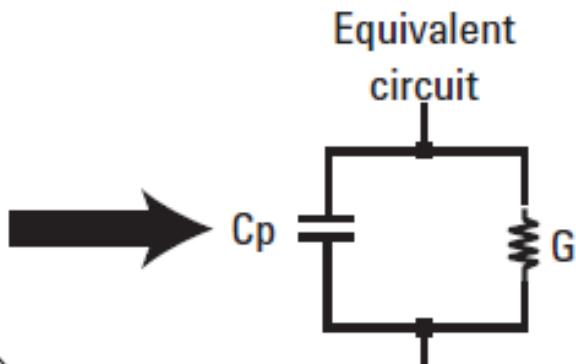
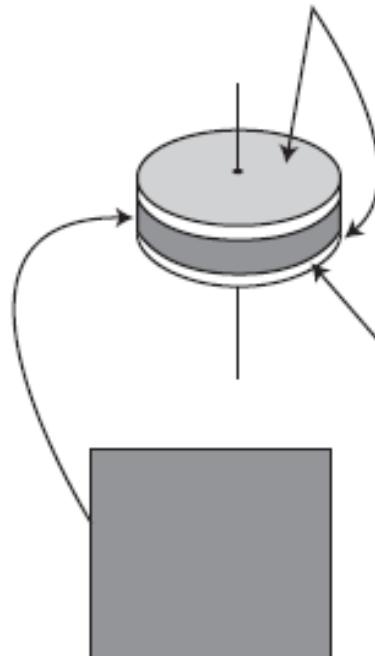


(magnetic
fixture also
available)



Parallel capacitor technique

Electrodes (Area = A)



$$Y = G + j\omega C_p$$
$$= j\omega C_0 \left(\frac{C_p}{C_0} - j \frac{G}{\omega C_0} \right)$$

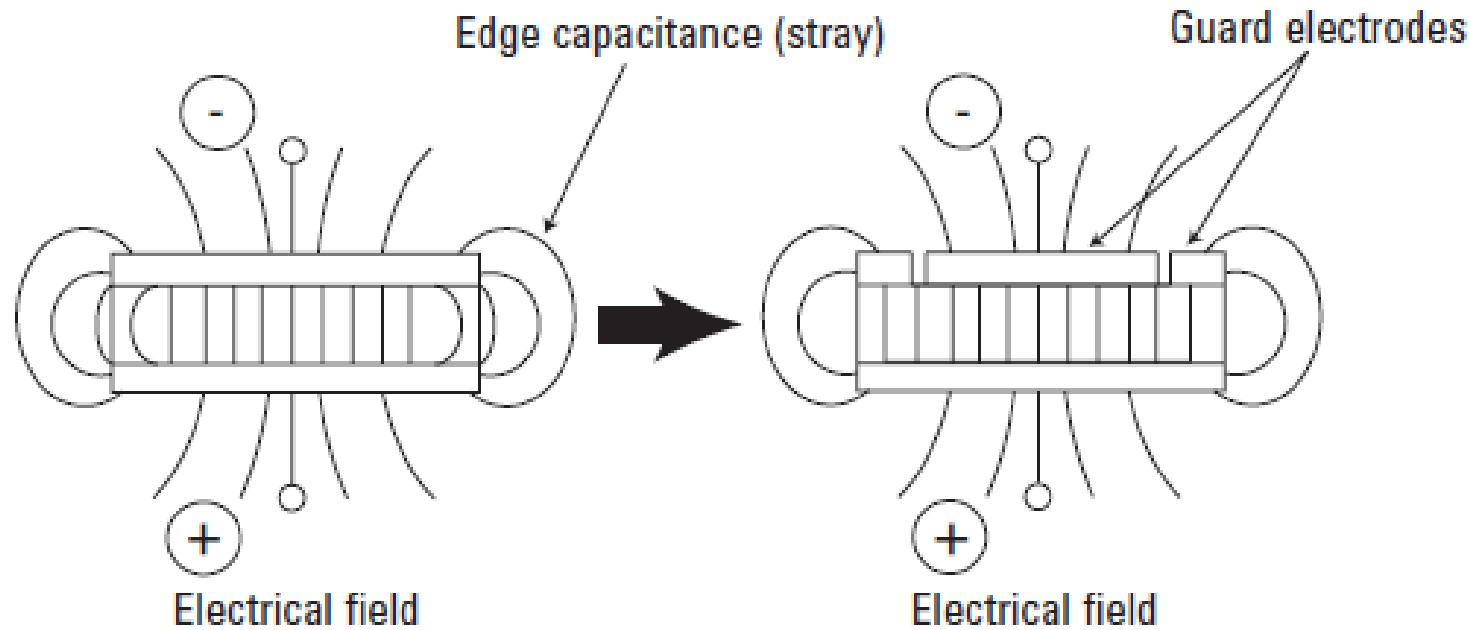
C_0 : Air capacitance

$$\epsilon_r^* = \left(\frac{C_p}{C_0} - j \frac{G}{\omega C_0} \right)$$

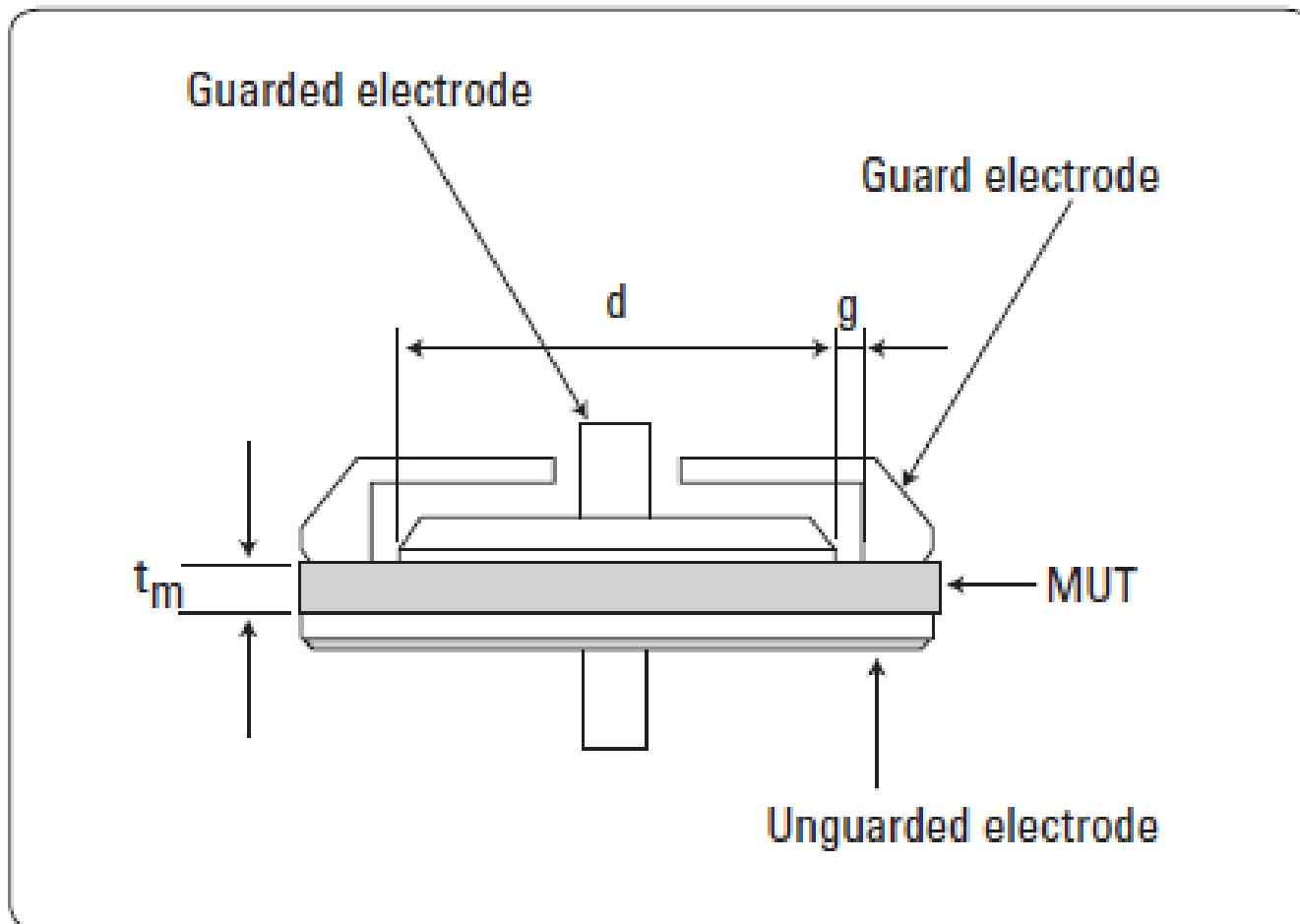
$$\epsilon_r' = \left(\frac{t^* C_p}{A^* \epsilon_0} \right)$$

$$\epsilon_r'' = \left(\frac{t}{\omega * R_p * A * \epsilon_0} \right)$$

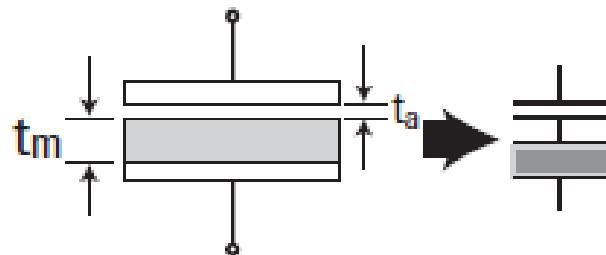
Effect of guard electrode



Contacting electrode method



Air gap effects



$$C_0 = \epsilon_0 \frac{A}{t_a} \quad \text{Capacitance of airgap}$$

$$C_x = \epsilon_x \epsilon_0 \frac{A}{t_m} \quad \text{Capacitance of dielectric material}$$

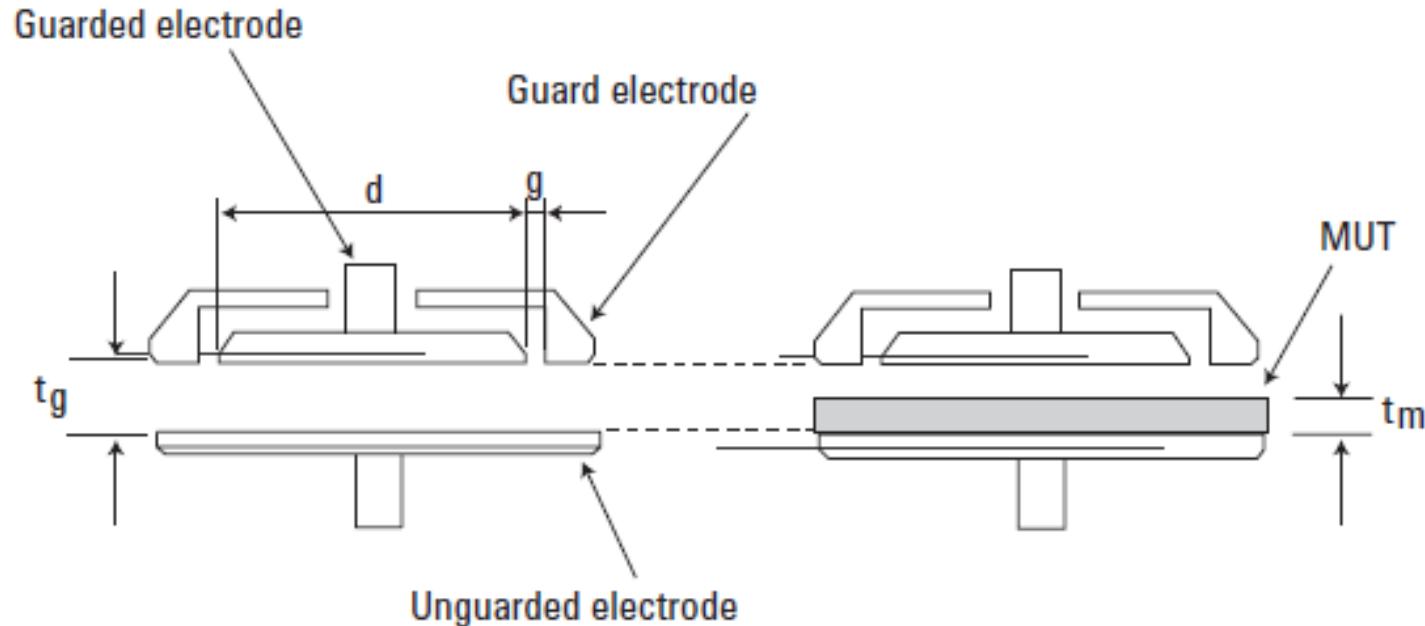
Measured capacitance

$$C_{\text{err}} = \frac{1}{\frac{1}{C_0} + \frac{1}{C_x}} = \epsilon_{\text{err}} \epsilon_0 \frac{A}{t_m + t_a}$$

Measured error
due to airgap

$$1 - \frac{\epsilon_{\text{err}}}{\epsilon_x} = \frac{\epsilon_x - 1}{\epsilon_x + \frac{t_m}{t_a}}$$

Non-contacting electrode method



Parallel plate measurement methods comparison



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Method	Accuracy	Application MUT	Operation
Contacting electrode	LOW	Solid material with a flat and smooth surface	1 measurement
Non-Contacting electrode	MEDIUM	Solid material with a flat and smooth surface	2 measurements
Thin film electrode	HIGH	Thin film electrode must be applied onto surfaces	1 measurement

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Coaxial probe system

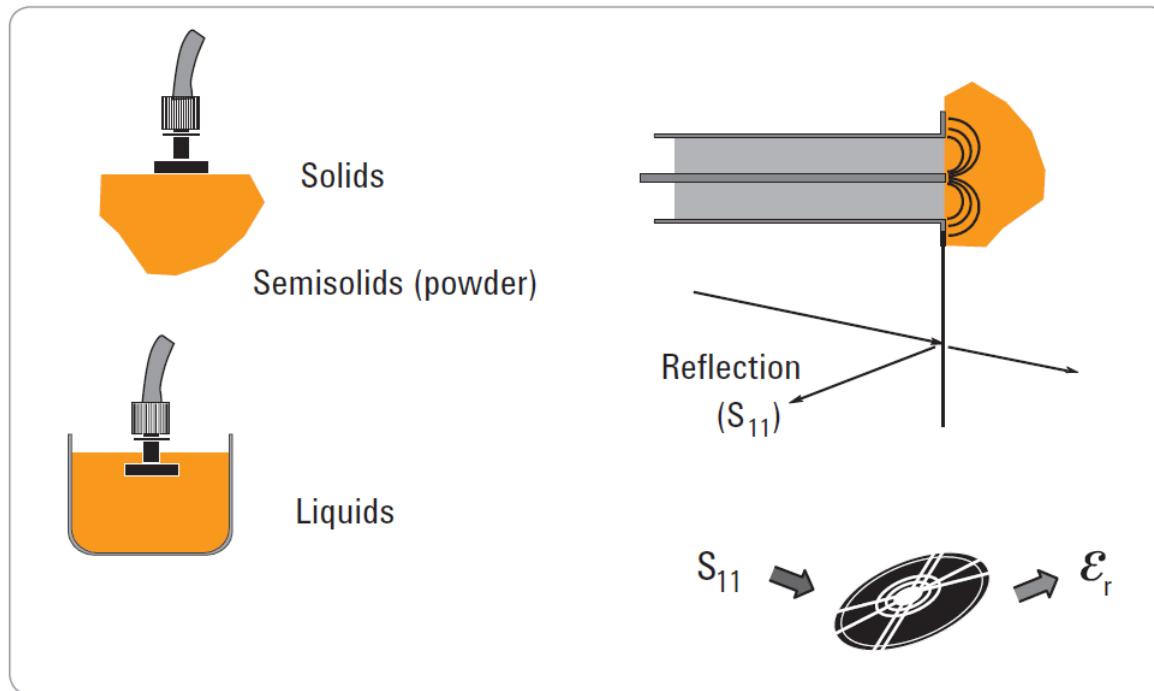


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Dielectric measurement setup for liquid using the coaxial probe method

Coaxial probe



Technique features:

- Broadband
- Simple and convenient (non destructive)
- Limited ϵ_r accuracy
- Limited $\tan \delta$ low loss resolution
- Best for liquids of semi-solids

Assumptions:

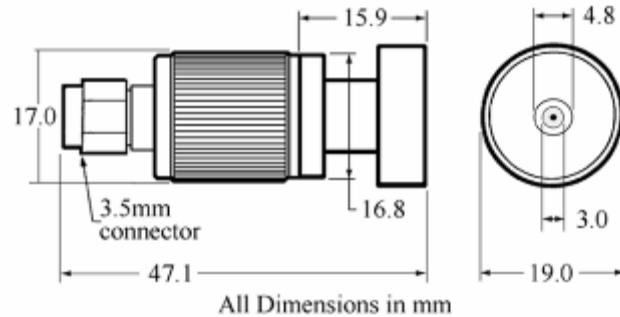
- Semi-infinite thickness
- Non-magnetic material
- Isotropic and homogeneous
- Flat surface
- No air gaps or bubbles

Three probe designs

1. High temperature probe



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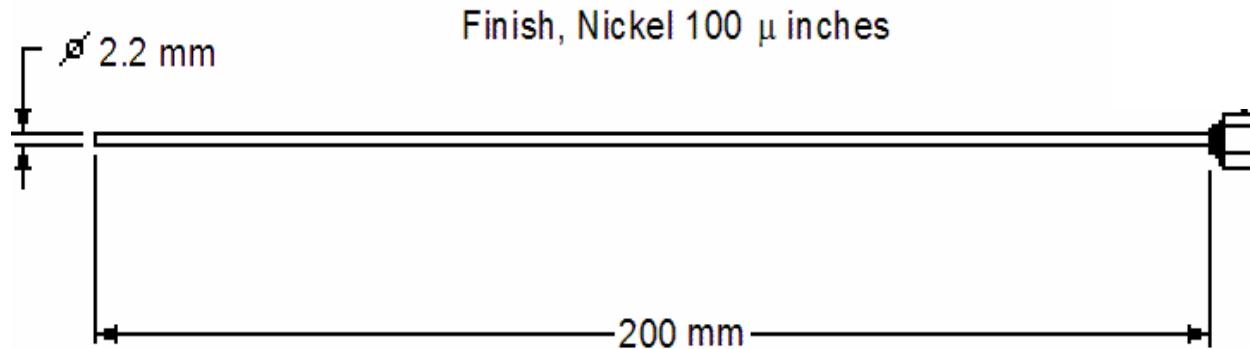


High Temperature Probe

- 0.200 – 20GHz (low end 0.01GHz with impedance analyzer)
- Withstands -40 to 200 degrees C
- Survives corrosive chemicals
- Flanged design allows measuring flat surfaced solids.

Three probe designs

2. Slim form probe



Slim Form Probe

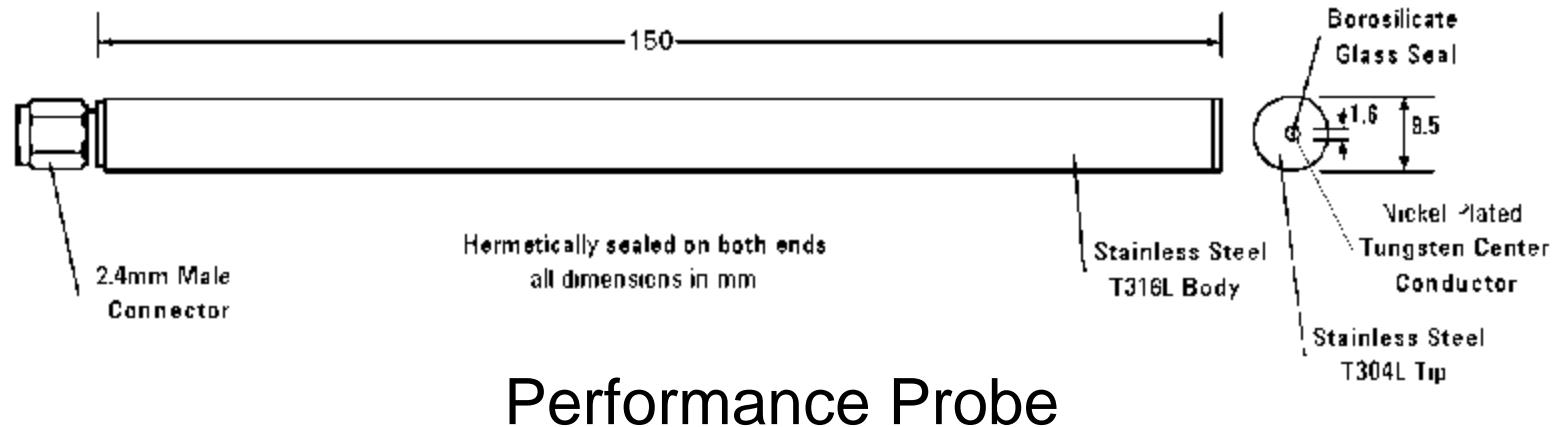
- 0.500 – 50GHz
- Low cost consumable design
- Fits in tight spaces, smaller sample sizes
- For liquids and soft semi-solids only

Three probe designs

3. Performance probe



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Performance Probe

Combines rugged high temperature performance with high frequency performance, all in one slim design.

- 0.500 – 50GHz
- Withstands -40 to 200 degrees C
- Hermetically sealed on both ends, OK for autoclave
- Food grade stainless steel



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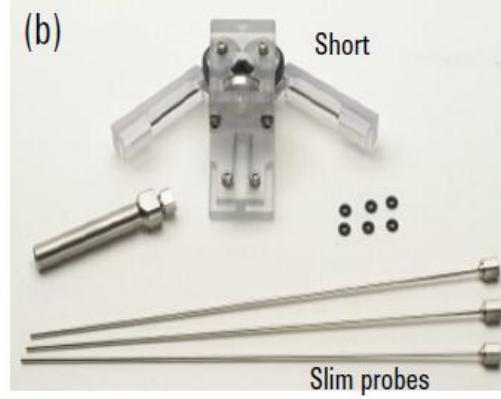
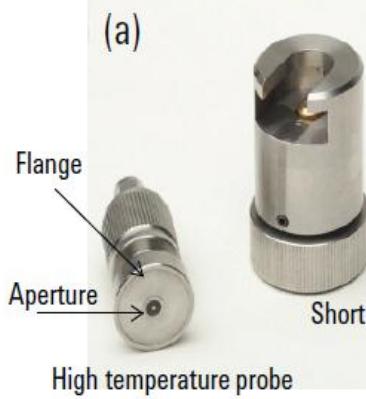
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Coaxial probe system



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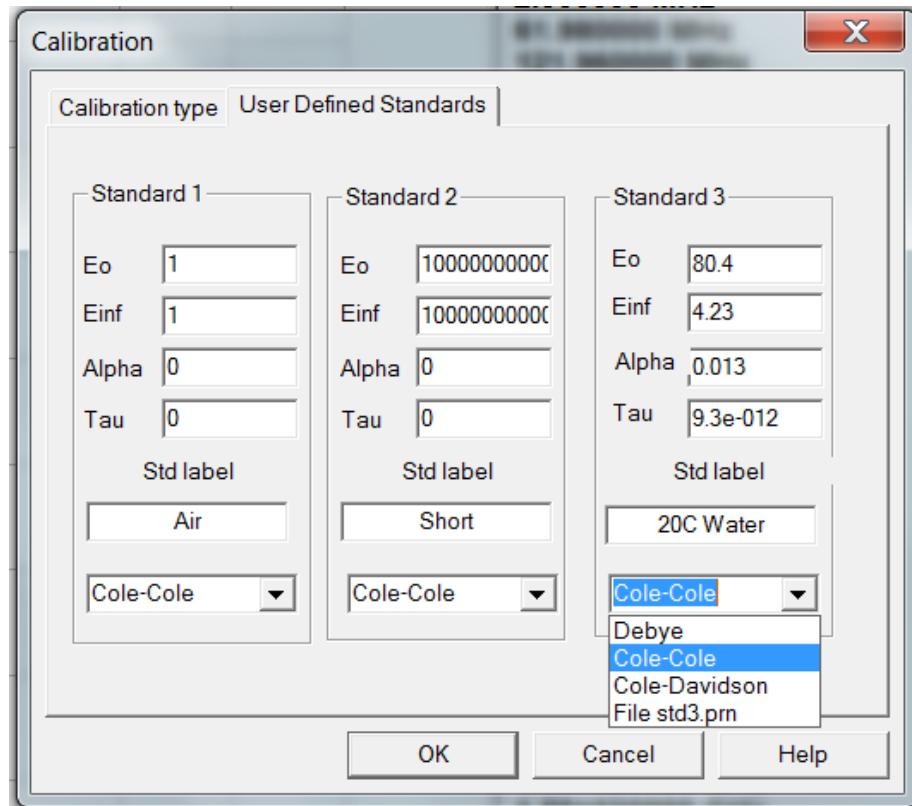
Calibration is required!!!



Coaxial probe system calibration



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Three standards:
Air, Short, Water

Air, Short, Load

User Defined Debye Cole

Cole Cole-Davidson

Permittivity Data

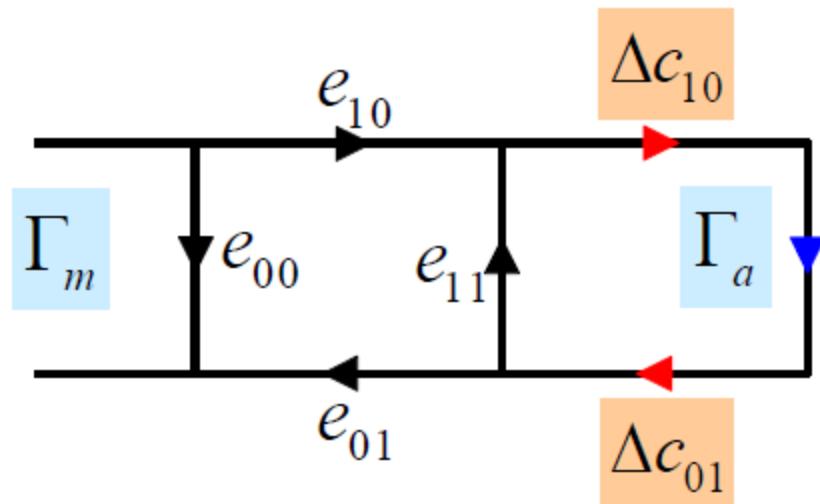


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Refresh calibration

If the perturbation is small, the change can be characterized by the measuring of a single calibration standard.



Γ_m = Measured S_{11}

Γ_a = Actual S_{11}

e_{00} = Directivity error

e_{11} = Source match error

$e_{10} e_{01}$ = Reflection tracking error

$\Delta c_{10} \Delta c_{01}$ = Perturbation term



Coaxial probe example data

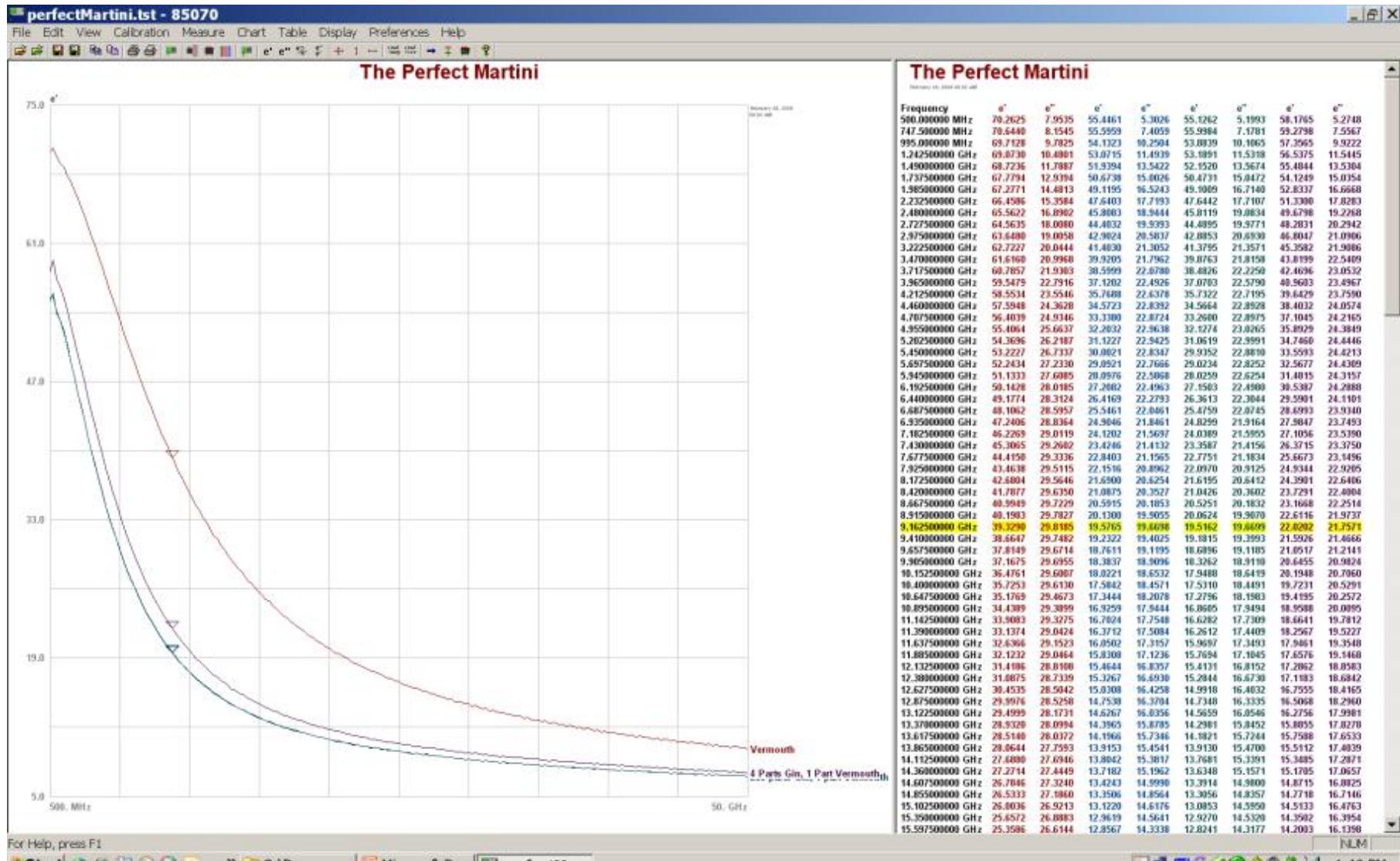
The perfect martini



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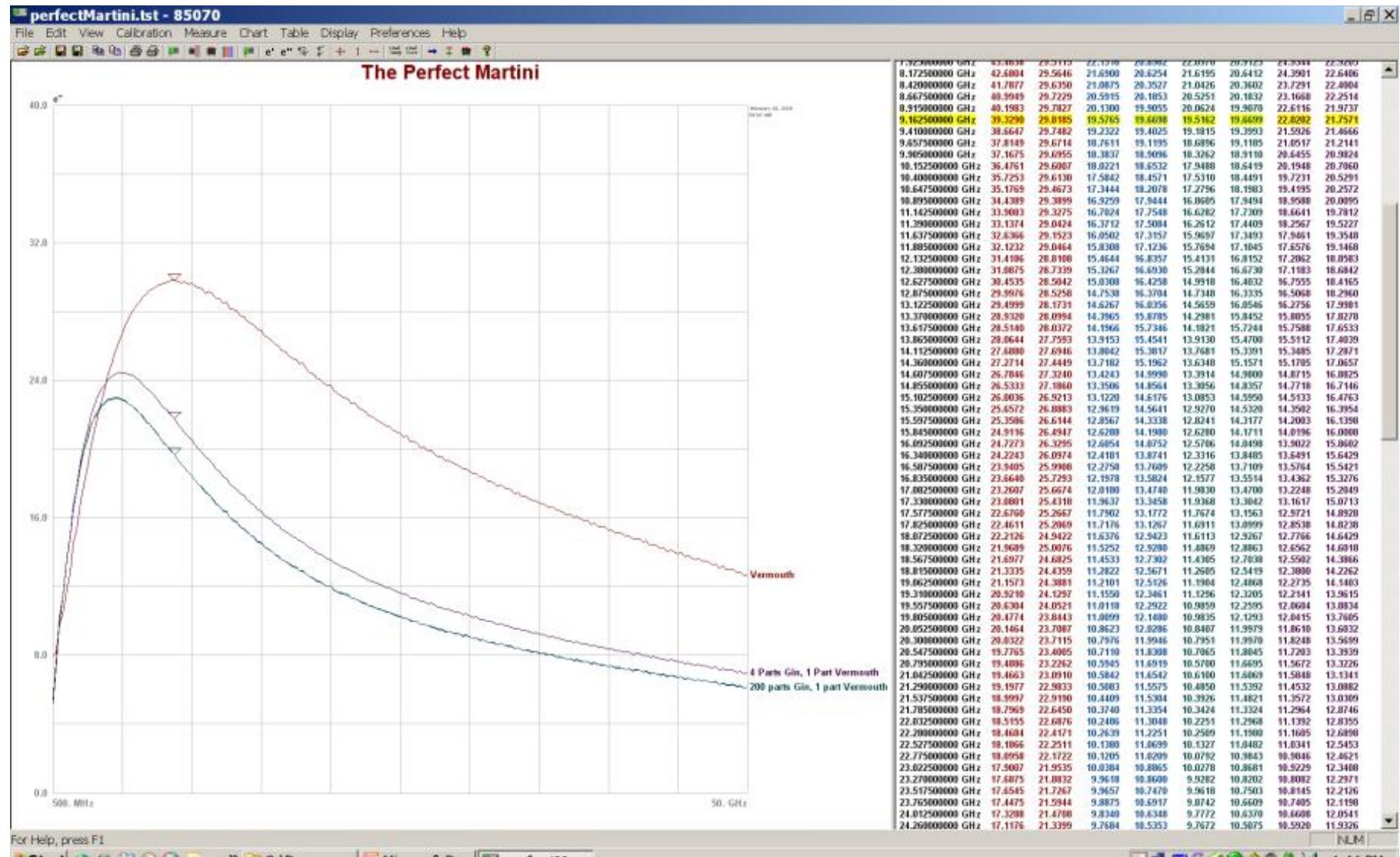
Coaxial probe example data



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Coaxial probe example data

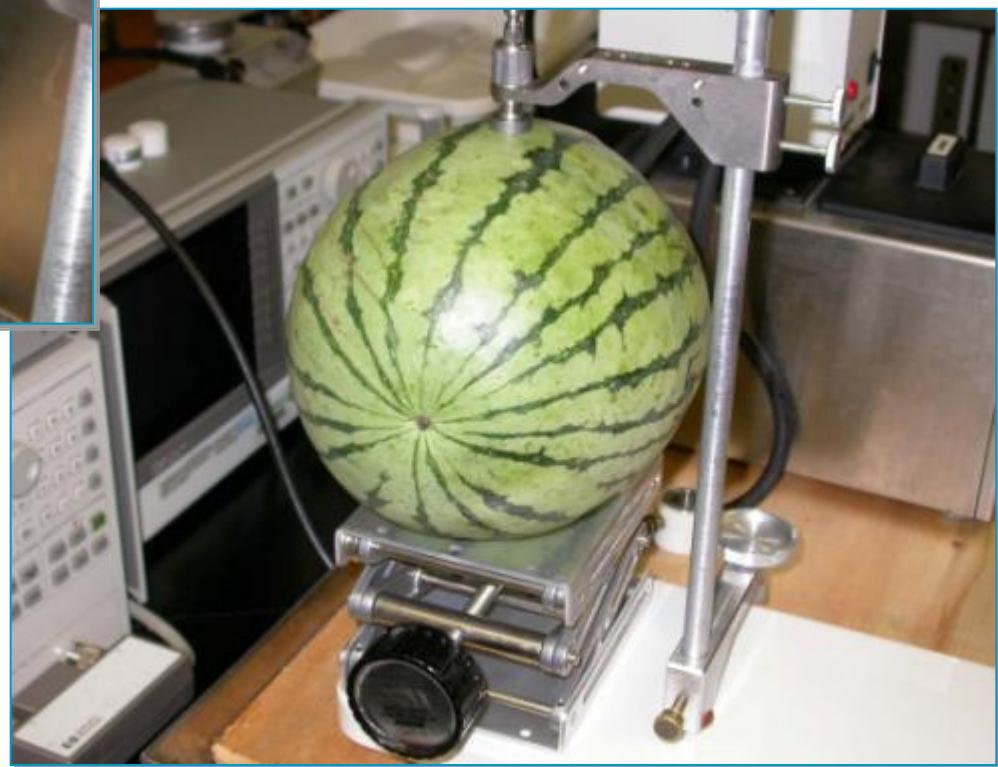


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USDA Fruit ripeness research



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CPAC Carbon nanotube research



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Agenda



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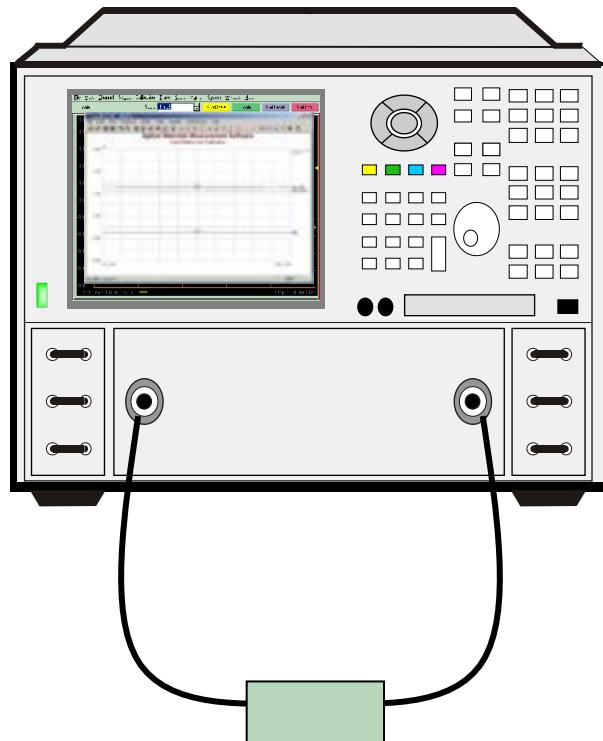
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- **Transmission line**
- Free space
- Resonant cavity

Transmission line system



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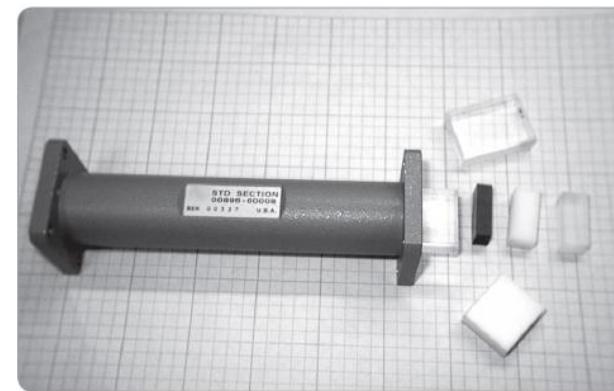
Network Analyzer



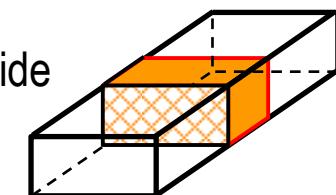
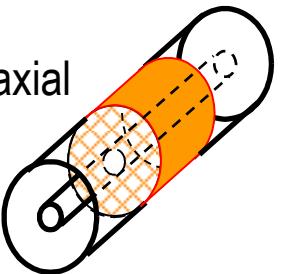
Sample holder
connected between coax cables



Coaxial



Waveguide

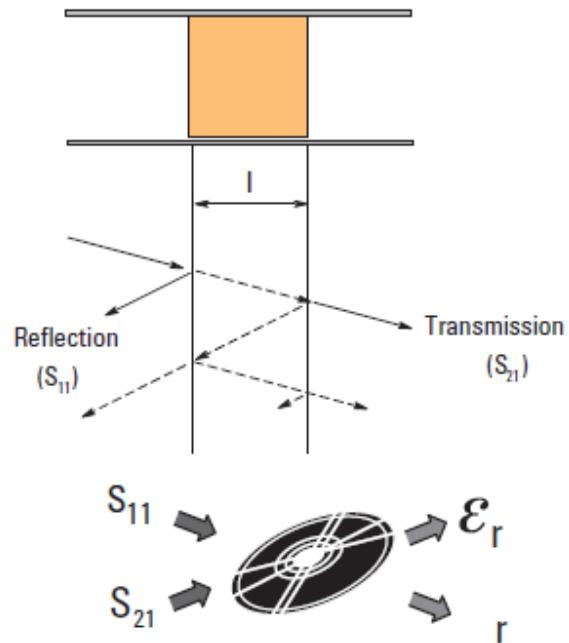
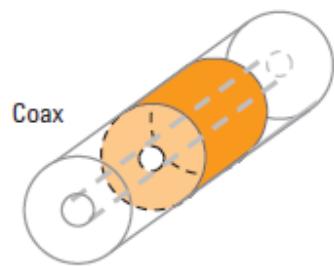
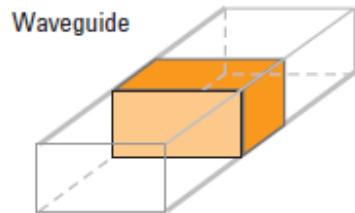


Calibration is required



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Transmission line



Technique features:

- Broadband (low freq. limited by practical sample length)
- Limited low loss resolution (depends on sample length)
- Measures magnetic materials
- Anisotropic materials can be measured in waveguide

Assumptions:

- Sample fill fixture across section
- No air gaps at fixture walls
- Smooth, flat faces, perpendicular to long axis
- Homogeneous

Transmission algorithms



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Algorithm	Measured S-parameters	Output
Nicolson-Ross	S11,S21,S12,S22	ϵ_r and μ_r
Precision (NIST)	S11,S21,S12,S22	ϵ_r
Fast	S21,S12	ϵ_r

(85071E also has three reflection algorithms)

Agenda



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Free space system

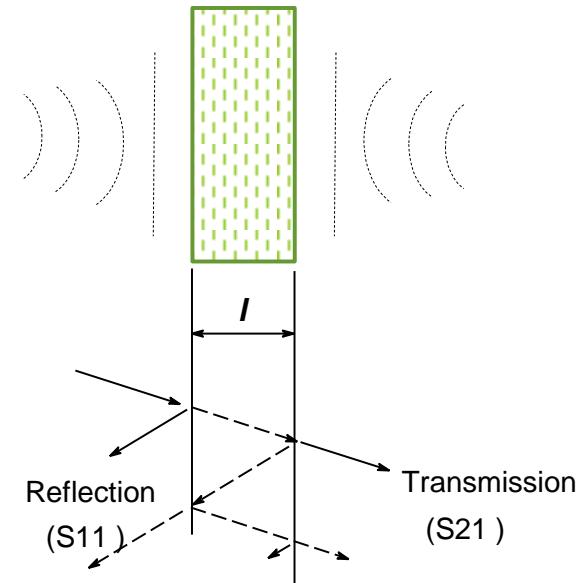
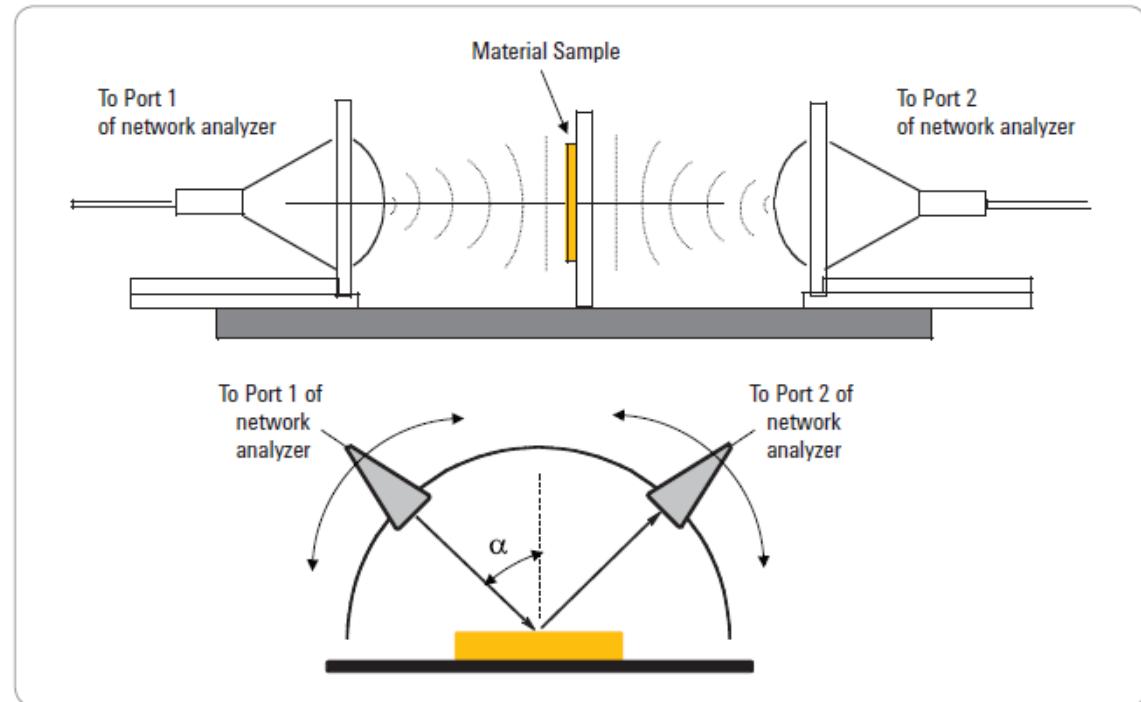


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Dielectric measurement setup for free space measurement

Free space system



Technique features:

- Non-contacting, non-destructive
- High frequency (low freq. Limited by practical sample size)
- Useful for high temperature
- Antenna polarization may be varied for anisotropic materials
- Measures magnetic materials

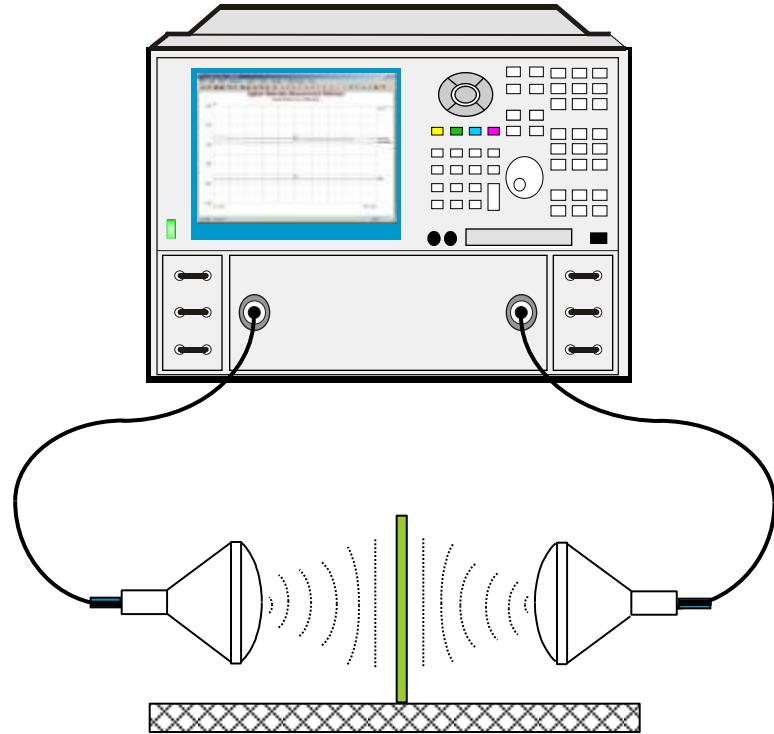
Assumptions:

- Flat parallel faced samples
- Sample in non-reactive region
- Beam spot is contained in sample
- Known thickness $> 20/360 \lambda$

Free space system requires calibration



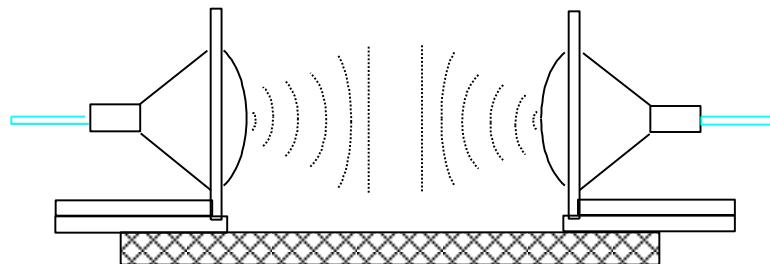
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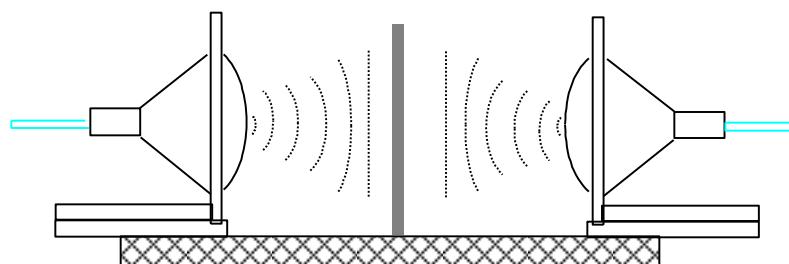
Before a measurement can be made, a calibration must be performed to remove systematic errors.

TRL calibration

Thru

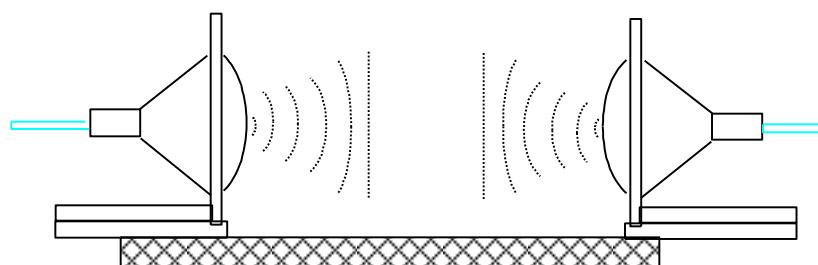


Reflect



Move the antenna away to compensate for the thickness of the short. Move it back for the next step.

Line

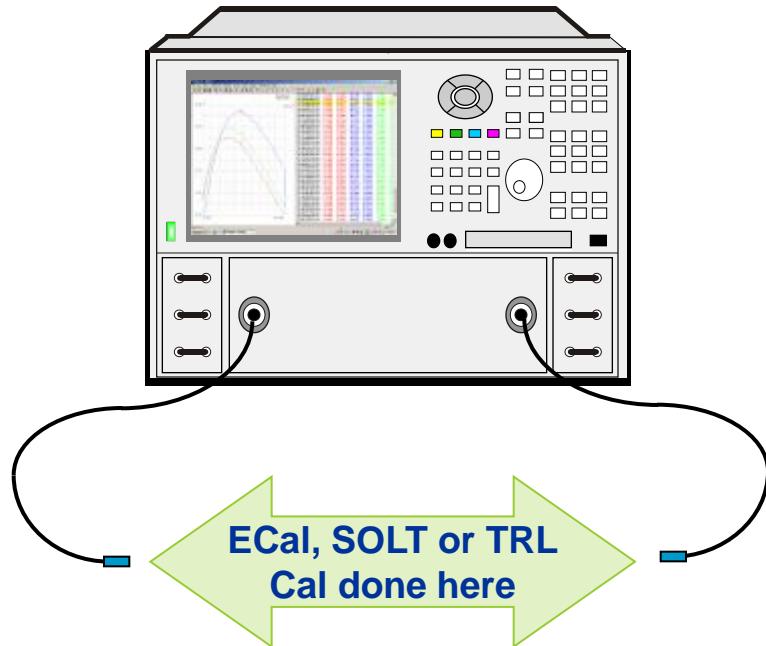


Move the antenna away on a quarter-wavelength and then back in the original position.

Two Tiered Calibration

1.

Two port calibration at waveguide or coax input into antennas removes errors associated with network analyzer and cables.

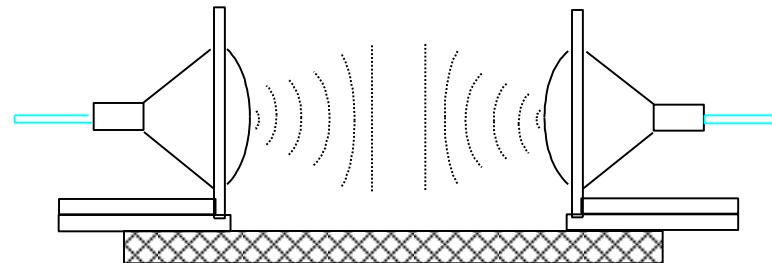


Two Tiered Calibration

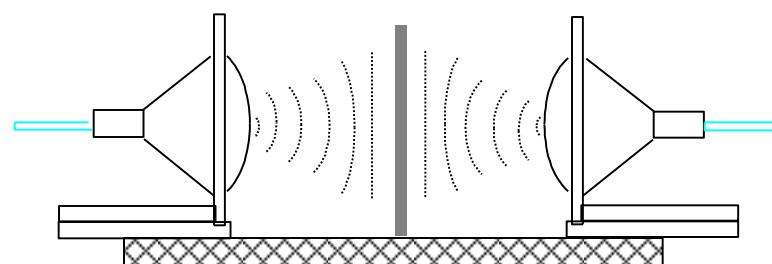
2.

Two additional free space calibration standards remove errors from antennas and fixture.

Line
(empty fixture)



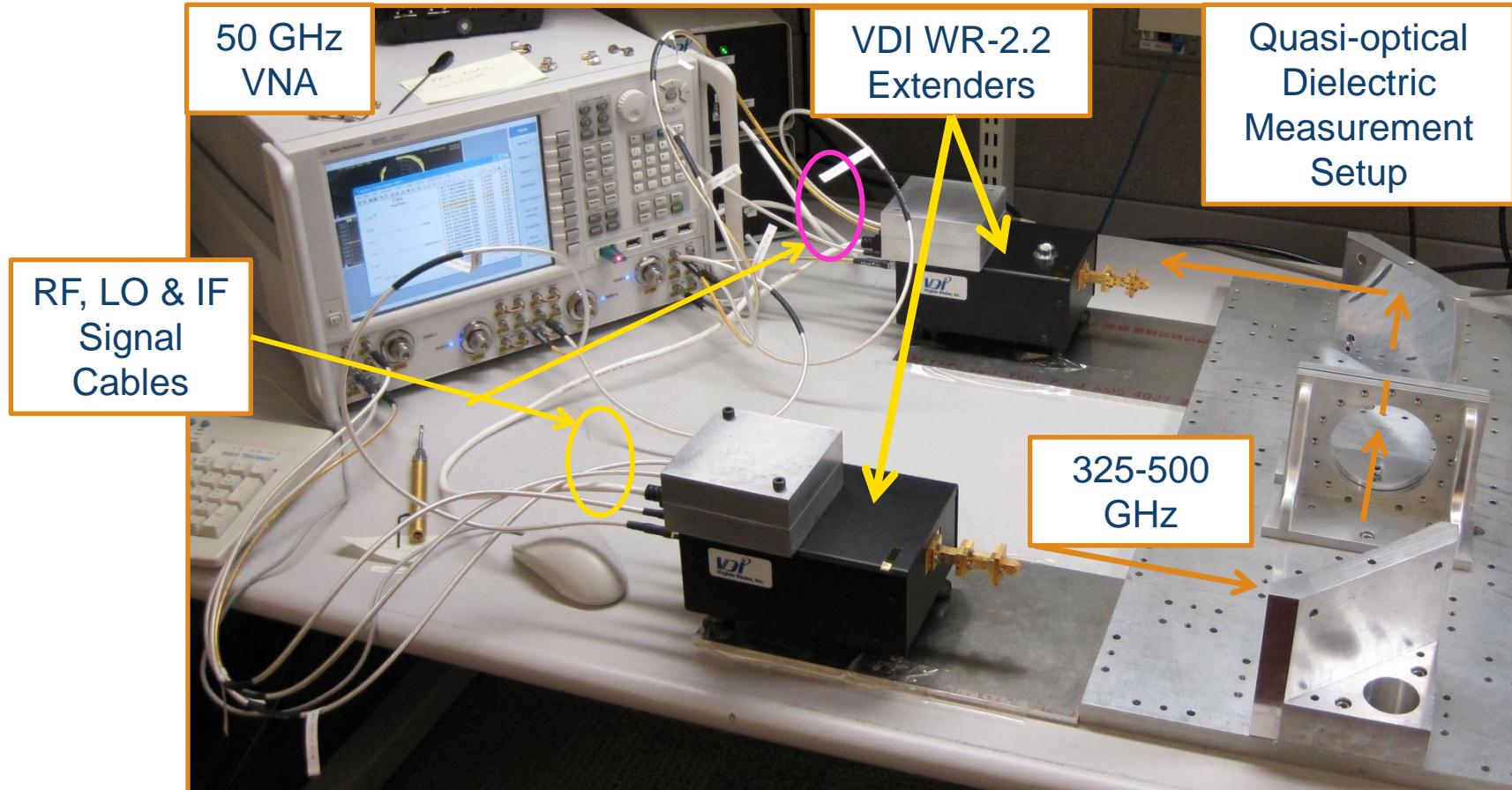
Reflect
(metal plate of
known thickness)



Quasi-optical VNA measurements

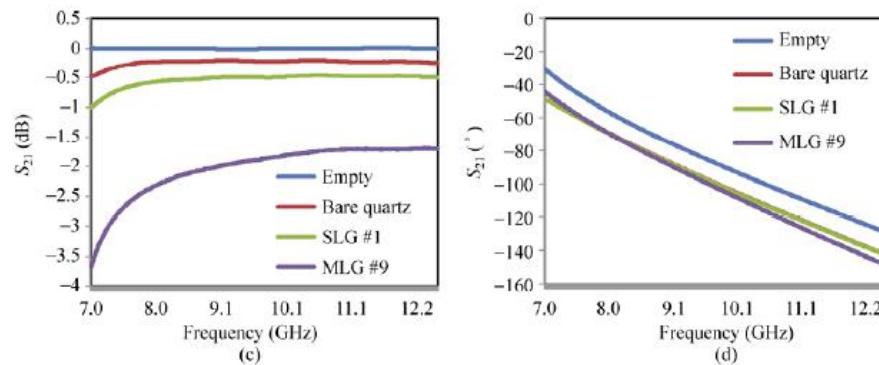
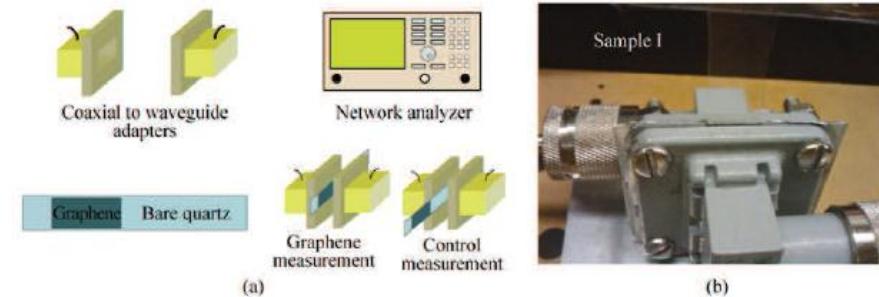


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- Quasi-optical dielectric measurements performed at Agilent

THz graphene characterization



Frequency extenders allow measurements to 1.1 THz

Measurement details

Frequency domain measurements of the absolute value of multilayer graphene (MLG) and single-layer graphene (SLG) sheet conductivity and transparency from DC to 1 THz

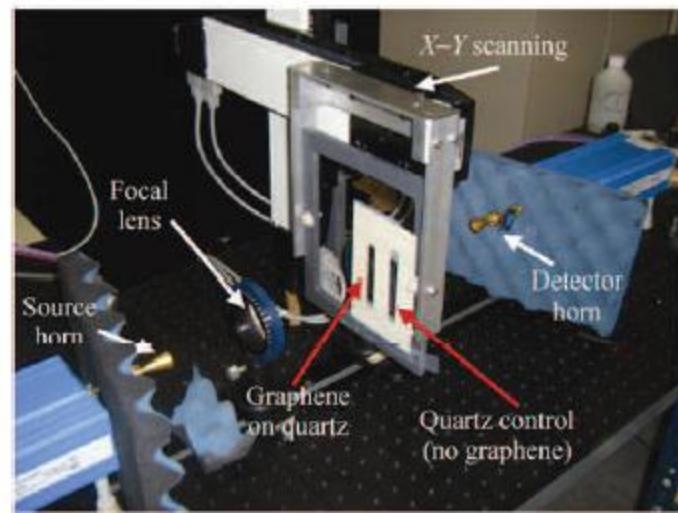


Figure 7 Picture of X-band imaging and quasi-optic transmission apparatus

Paper details

Terahertz Graphene Optics

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Agenda



Unlocking Measurement Insights for 75 Years

- Introduction
- Permittivity and permeability
- Measurement techniques and systems
- Parallel plate
- Coaxial probe
- Transmission line
- Free space
- **Resonant cavity**

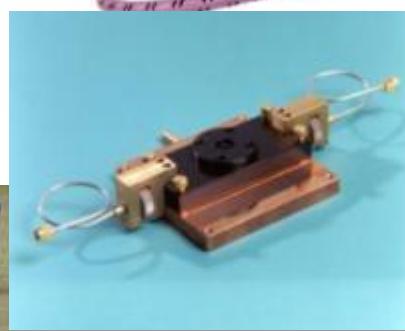
Resonant cavity system



Agilent Split Cylinder Resonator
IPC TM-650-2.5.5.5.13



ASTM 2520 Waveguide
Resonators



Split Post Dielectric
Resonators from QWED

Resonant cavity technique



f_c = Resonant Frequency of Empty Cavity

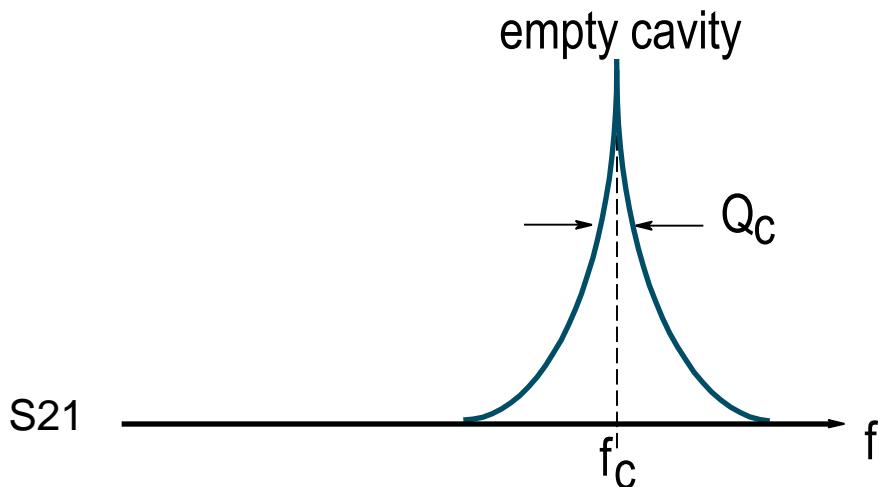
f_s = Resonant Frequency of Filled Cavity

Q_c = Q of Empty Cavity

Q_s = Q of Filled Cavity

V_s = Volume of Empty Cavity

V_c = Volume of Sample



S21



$$\begin{aligned}\varepsilon'_r &= 1 + \frac{V_c(f_c - f_s)}{2V_s f_s} = \\ \varepsilon''_r &= \frac{V_c}{4V_s} \left(\frac{1}{Q_s} - \frac{1}{Q_c} \right)\end{aligned}$$

ASTM 2520

Resonant cavity technique

f_c = Resonant Frequency of Empty Cavity

f_s = Resonant Frequency of Filled Cavity

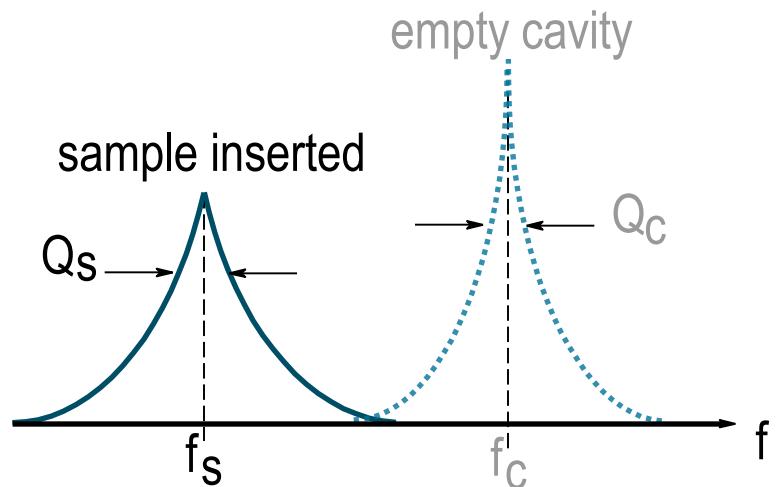
Q_c = Q of Empty Cavity

Q_s = Q of Filled Cavity

V_s = Volume of Empty Cavity

V_c = Volume of Sample

S21



$$\begin{aligned}\varepsilon'_r &= 1 + \frac{V_c(f_c - f_s)}{2V_s f_s} = \\ \varepsilon''_r &= \frac{V_c}{4V_s} \left(\frac{1}{Q_s} - \frac{1}{Q_c} \right)\end{aligned}$$

ASTM 2520

Resonant cavity technique

f_c = Resonant Frequency of Empty Cavity

f_s = Resonant Frequency of Filled Cavity

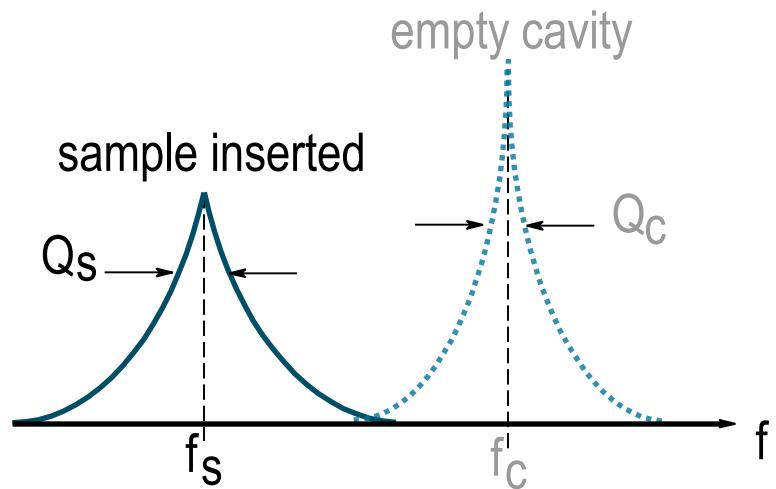
Q_c = Q of Empty Cavity

Q_s = Q of Filled Cavity

V_c = Volume of Empty Cavity

V_s = Volume of Sample

S21



$$\epsilon'_r = 1 + \frac{V_c (f_c - f_s)}{2V_s f_s} =$$
$$\epsilon''_r = \frac{V_c}{4V_s} \left(\frac{1}{Q_s} - \frac{1}{Q_c} \right)$$

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Resonant cavity technique

f_c = Resonant Frequency of Empty Cavity

f_s = Resonant Frequency of Filled Cavity

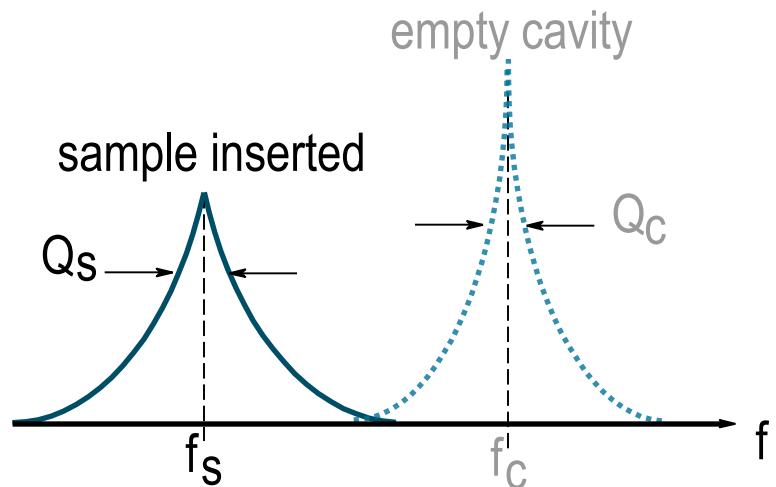
Q_c = Q of Empty Cavity

Q_s = Q of Filled Cavity

V_s = Volume of Empty Cavity

V_c = Volume of Sample

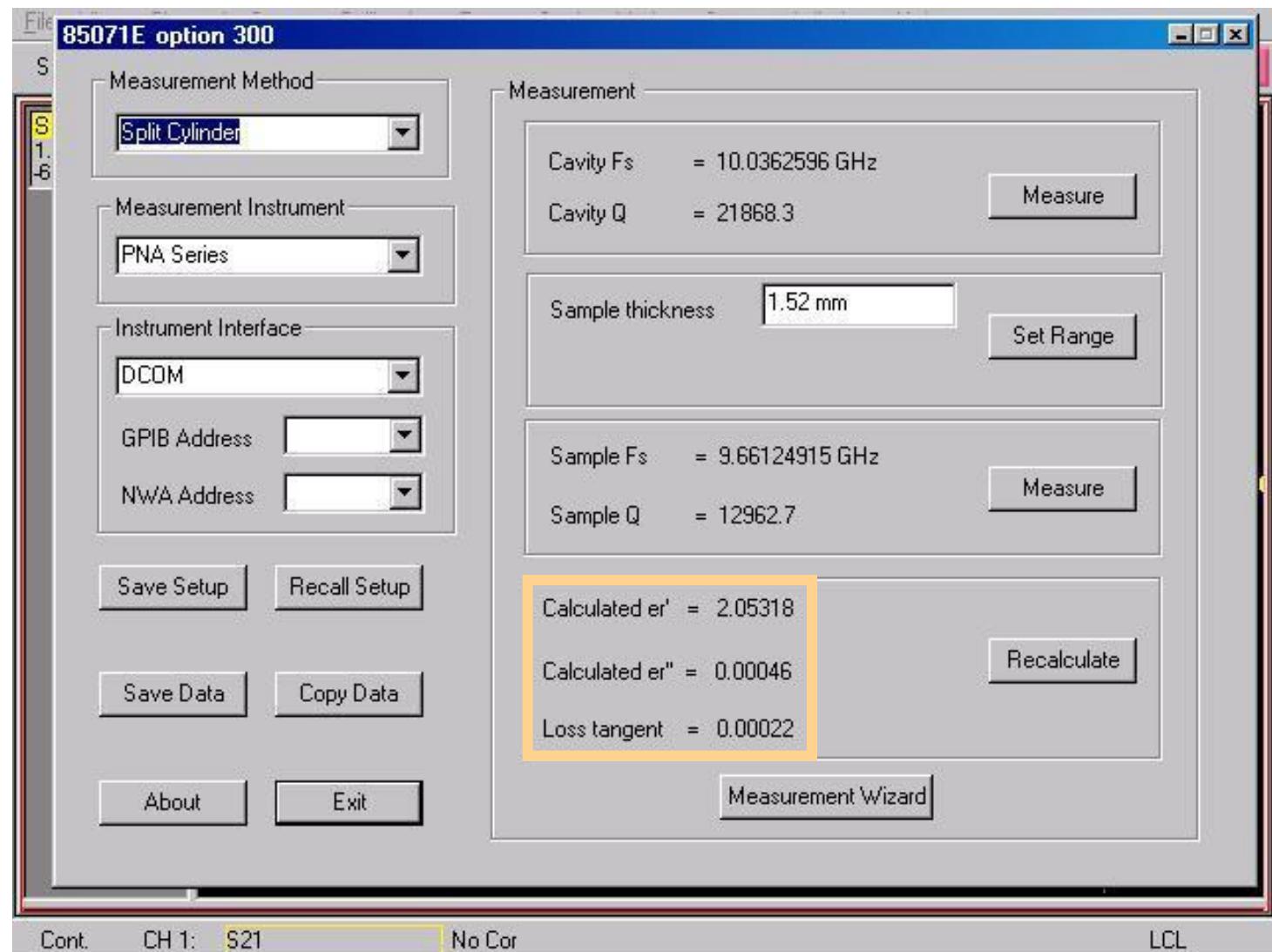
S21



$$\begin{aligned}\varepsilon'_r &= 1 + \frac{V_c(f_c - f_s)}{2V_s f_s} = \\ \varepsilon''_r &= \frac{V_c}{4V_s} \left(\frac{1}{Q_s} - \frac{1}{Q_c} \right)\end{aligned}$$

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Resonant cavity example data



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Resonant vs. broadband transmission



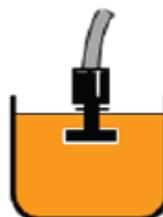
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	Resonant	Broadband
Low Loss materials	Yes ϵ_r'' resolution $\leq 10^{-4}$	No ϵ_r'' resolution $\geq 10^{-2}$ - 10^{-3}
Thin Films and Sheets	Yes 10GHz sample thickness $< 1\text{mm}$	No 10GHz optimum thickness ~ 5-10mm
Calibration Required	No	Yes
Measurement Frequency Coverage	Discrete Frequencies	Broadband or Banded

Summary technique and strengths

Coaxial Probe

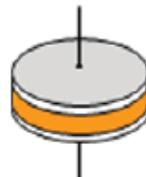
ϵ_r



Broadband, convenient, non-destructive
Best for lossy MUTs; liquids and semi-solids

Parallel Plate

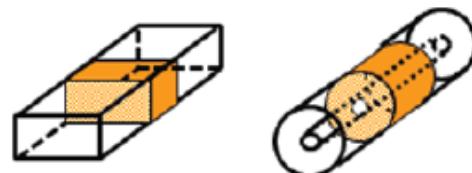
ϵ_r



Accurate
Best for low frequencies; thin, flat sheets

Transmission Line

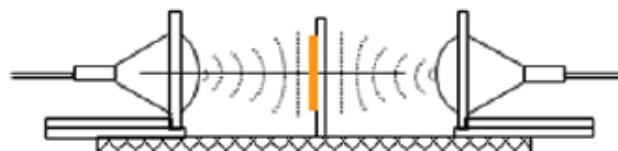
ϵ_r and μ_r



Broadband
Best for lossy to low loss MUTs;
machineable solids

Free Space

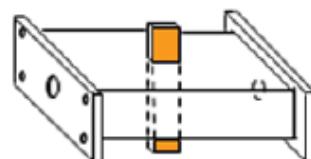
ϵ_r and μ_r



Broadband; Non-contacting
Best for flats sheets, powders, high temperatures

Resonant Cavity

ϵ_r



Single frequency; Accurate
Best for low loss MUTs; small samples

References



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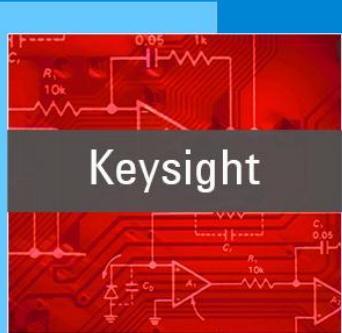


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Thank you!



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