How to Validate BLVDS SER/DES Signal Integrity Using an Eye Mask

The following application note contains information that will help you validate signal quality on a BLVDS SER/DES link. How to capture an eye pattern, how to generate an eye mask, and how to validate signal quality are all explained in detail in this application note. The National DS92LV1023 and DS92LV1224 are used as examples, but this procedure may be used with other BLVDS SER/DES devices. The 1023/ 1224 chipset operates between 40 and 66 MHz and serializes a 10-bit data bus. Datasheets for the DS92LV1023 and DS92LV1224, as well as information on other LVDS and Bus LVDS devices, can be found on National's website at http://www.LVDS.National.com.

How to Capture a Serialized BLVDS Eye Pattern

Capturing a Low Voltage Differential Signal (LVDS) is simple as long as you follow a few general guidelines. First, make sure you are operating within the specifications for the National Semiconductor Application Note 1217 James Chang March 2002



DS92LV1023 and DS92LV1224. In order to see the full twelve bits (10 data bits and 2 embedded clock bits), you must trigger on an external clock source. In this scenario, we will use the same source as the TCLK input for the DS92LV1023. Pseudo random data should be applied to the transmitter data input pins along with clock from a data generator or other source. The display on the oscilloscope should also be set to Variable Persistence Mode. In order to accurately probe the LVDS signal, a high bandwidth (>1 GHz) and high impedance (>100k Ω) differential probe should be used. In the example shown below, a Tektronix TEKP6248 differential probe was used.

The Ideal Start and Stop Bit Position

After using the recommendations above, you should be able to capture an eye pattern that looks like Figure 1. The bits have been labeled for clarity.

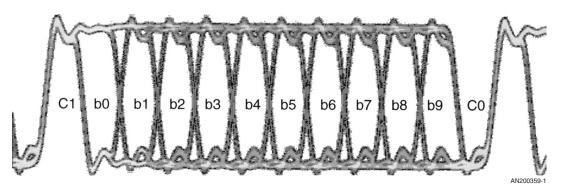


FIGURE 1. BLVDS Serialized Payload Eye Pattern with PRBS Data

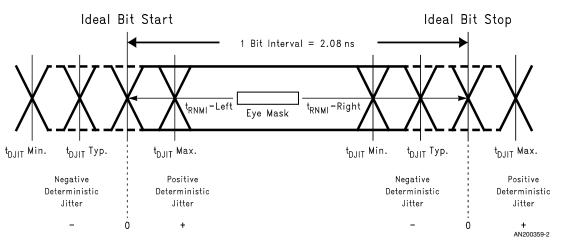


FIGURE 2. Single BLVDS Data Bit Drawing

Figure 2 shows all parameters involved when isolating a single bit from the data stream. Not all parameters will apply, and in order to determine which parameters will apply, the ideal start and stop bit positions need to be calculated. By determining the ideal bit positions, we can determine if the start and stop positions for the captured signal are early, late, or ideal. For this example, we will use a 40 MHz clock with a

period (T) of 25.0ns. There are a few guidelines to remember when making start and stop bit calculations. For one, all calculations should be made from the first clock (C1) edge. This should be considered time zero for the period. Ideally, each bit is T/12 wide, or in this case, 25.0ns/12 = 2.083ns wide per bit. The bits start and stop in increments of 2.083ns from the C1 rising edge. Applying this, we arrive at *Table 1*.

TABLE 1. Ideal Start and Stop I	Bit	Positions
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D:4 #	Bit	Ideal Bit	Ideal Bit	Ideal Bit
Bit #	Desc.	Start (ns)	Center (ns)	Stop (ns)
1	C1	0.000	1.040	2.083
2	0	2.083	3.123	4.166
3	1	4.166	5.206	6.249
4	2	6.249	7.289	8.332
5	3	8.332	9.372	10.415
6	4	10.415	11.455	12.498
7	5	12.498	13.538	14.581
8	6	14.581	15.621	16.664
9	7	16.664	17.704	18.747
10	8	18.747	19.787	20.830
11	9	20.830	21.870	22.913
12	C0	22.913	23.953	24.996
	Condition	ns: Frequency = 40 MHz;	Period (T) = 25.0 ns	

It is easiest to tell if a bit's start or stop position is early or late by using the vertical cursor function on the oscilloscope. One vertical cursor should be placed on the C1 rising edge (time zero) and another should be placed at the crossing point of the isolated bit. Use the 0V (differential) crossing and the center of the distribution to set the first cursor on the clock transition center (C0 to C1 rising edge). Move the second cursor to the ideal location (a delta of 2.083ns in this example) for the ideal start location of the first data bit. If the waveform on the oscilloscope transitions before the cursor, the bit is early. If aligned, it is at the ideal location, and if after, it is late. Minor offsets are expected due to PLL constraints, clock purity, and external noise.

Adjusting to Ideal Bit Positions and Placing the Eye Mask

In placing the leading edge of an Eye Mask, the three important parameters are the ideal bit start position, $t_{\text{RNMI-LEFT}}$, and t_{DJIT} typical. Because the eye mask may not be perfectly centered within the ideal bit (it may actually be on the early side of the bit, for example), only the worst-case (leading edge) $t_{\text{RNMI-LEFT}}$ min. is specified. The noise margin on the trailing side of the measured bit is larger. Looking at the t_{DJIT} specification from the DS92LV1023 datasheet, we see that, typically, the start and stop bit positions are 80ps early (designated by a negative sign). This is depicted in Figure 3.

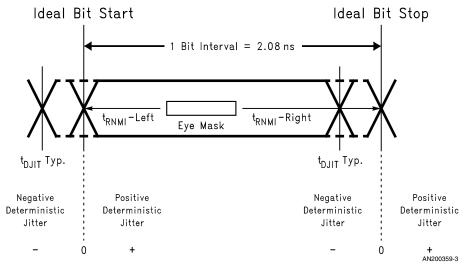


FIGURE 3. Parameter Relationships to Generate the Eye Mask

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Looking at the DS92LV1224 datasheet at 40 MHz, the worst-case noise margin ($t_{\rm RNM}$ min.) is 450ps. The trailing edge of the eye mask has a larger noise margin, and thus we use the typical noise margin value of 730ps for this measurement. This information is all that is required to set the eye mask. However, because the noise margin specification is measured to the ideal start and stop bit positions, it needs to be budgeted against transmitter deterministic jitter ($t_{\rm DJIT}$ max.), intersymbol interference (ISI), and available margin. To ensure error free operation, no hits should be observed in the eye mask.

Leading edge: $t_{RNMI-LEFT}$ Min. – t_{DJIT} typ. = 450ps – (–80ps) = 530ps

Trailing edge: $t_{RNMI-RIGHT}$ Typ. + t_{DJIT} typ. = 730ps + (-80ps) = 650ps

Figure 4 shows a single scope captured bit with a properly placed eye mask. The captured signal has been outlined for clarity.

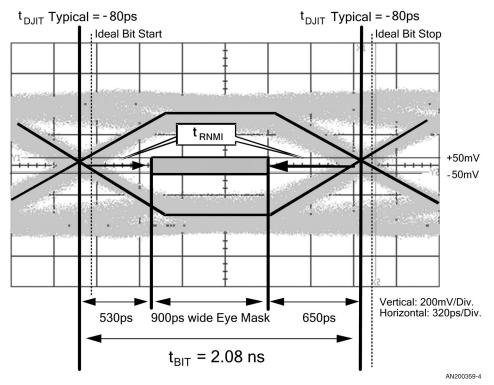


FIGURE 4. Captured Eye Pattern with Properly Placed Eye Mask

So on the scope, the leading edge of the mask should be placed 530ps after the middle of the observed crossing. And the trailing edge of the mask should be placed 650ps before the next observed crossing. Thus, in this example, the Eye Mask is about 900ps wide. Typically, the strobe window for the DS92LV1224 is about 100ps wide. However, the position of this window shifts due to bit order, process, temperature, and frequency. Therefore, the 900ps eye mask is a conservative mask that includes a four-sigma margin.

The minimum and maximum specifications for t_{DJIT} are not necessary to place the eye pattern mask but can be used to determine interoperability between Serializers and Deserializers. For example, it would not be advisable to use a Serializer with a typical deterministic jitter of 0ps and a maximum deterministic jitter of +300ps with a Deserializer with a noise margin of 300ps; the deterministic jitter from the Serializer would use up all of the receiver's noise margin, leaving no margin for inevitable interconnect loading (ISI).

Validating Signal Quality

For reliable link operation, the opening of the eye pattern should be outside of the eye mask. We have already determined where to place the leading and trailing edge of the eye mask, so now we need to draw the actual height (amplitude)

for the eye mask. The amplitude of the eye mask is simply the differential receiver's input threshold voltages (V_{TH} and V_{TL}), which are typically $\pm 50 \text{mV}$. In some cases, a $\pm 100 \text{mV}$ level is used to make the verification measurement more conservative. As long as the captured eye opening stays out of the eye mask, the chipset should be able to operate error free. Note that this validation is done at the receive end of the link, after the interconnect loading effects have occurred.

Applicable Devices

Current and future datasheet noise margin specifications will not account for transmitter deterministic jitter. All current and future parts will measure eye mask corners to ideal bit positions. The following table lists all current (Q1 CY2002) devices where this application note applies. In addition, all future product datasheets will separate the left ($t_{\rm RNMI-LEFT}$) and right ($t_{\rm RNMI-RIGHT}$) Noise Margin specifications and will use this methodology to validate signal quality.

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TABLE 2. Current (Q1 CY2002) Applicable Parts

Part Number	Description
DS92LV1212A	10-bit Deserializer
DS92LV1224	10-bit Deserializer
DS92LV1260	Six Channel 10-bit Deserializer
DS92LV16	16-bit Transceiver
SCAN921226	10-bit Deserializer with SCAN
SCAN921260	Six Channel 10-bit Deserializer with
	SCAN

Note 1: The DS92LV16 is a 16-bit Transceiver; calculations used must be adjusted accordingly.

First generation devices measured noise margin to the maximum t_{DJIT} limit. These parts are listed in *Table 3*. Thus, for these devices, noise margin is a chipset specification and is allocated solely to the interconnect. The measurement location of t_{RNM} was changed to allow inter-operation calculations between different devices within the family (i.e. 10-bit payload).

TABLE 3. Past Non-Applicable Parts

Part Number	Description
DS92LV1210	10-bit Deserializer
DS92LV1212	10-bit Deserializer

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