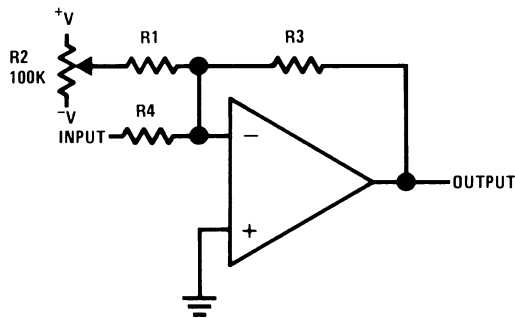


Universal Balancing Techniques

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IC op amps are widely accepted as a universal analog component. Although the circuit designs may vary, most devices are functionally interchangeable. However, offset voltage balancing remains a personality trait of the particular amplifier design. The techniques shown here allow offset voltage balancing without regard to the internal circuitry of the amplifier.



$$R1 = 2000 R3 \parallel R4$$

$$R4 \parallel R3 \leq 10 \text{ k}\Omega$$

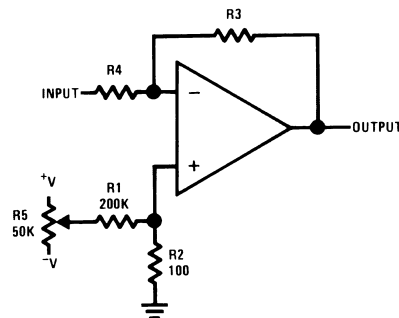
$$\text{RANGE} = \pm V \left(\frac{R3 \parallel R4}{R1} \right)$$

FIGURE 1. Offset Voltage Adjustment for Inverting Amplifiers Using 10 kΩ Source Resistance or Less

The circuit shown in *Figure 1* is used to balance out the offset voltage of inverting amplifiers having a source resistance of 10 kΩ or less. A small current is injected into the summing node of the amplifier through R_1 . Since R_1 is 2000 times as large as the source resistance the voltage at the arm of the pot is attenuated by a factor of 2000 at the summing node. With the values given and $\pm 15\text{V}$ supplies the output may be zeroed for offset voltages up to $\pm 7.5 \text{ mV}$.

If the value of the source resistance is much larger than 10 kΩ, the resistance needed for R_1 becomes too large. In this case it is much easier to balance out the offset by supplying a small voltage at the non-inverting input of the amplifier. *Figure 2* shows such a scheme. Resistors R_1 and R_2 divide the voltage at the arm of the pot to supply a $\pm 7.5 \text{ mV}$ adjustment range with $\pm 15\text{V}$ supplies.

This adjustment method is also useful when the feedback element is a capacitor or non-linear device.

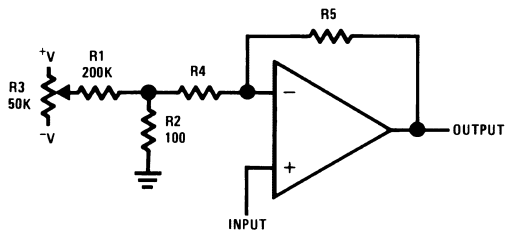


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$$\text{RANGE} = \pm V \left(\frac{R2}{R1} \right)$$

FIGURE 2. Offset Voltage Adjustment for Inverting Amplifiers Using Any Type of Feedback Element

This technique of supplying a small voltage effectively in series with the input is also used for adjusting non-inverting amplifiers. As is shown in *Figure 3*, divider R_1 , R_2 reduces the voltage at the arm of the pot to $\pm 7.5 \text{ mV}$ for offset adjustment. Since R_2 appears in series with R_4 , R_2 should be considered when calculating the gain. If R_4 is greater than 10 kΩ the error due to R_2 is less than 1%.



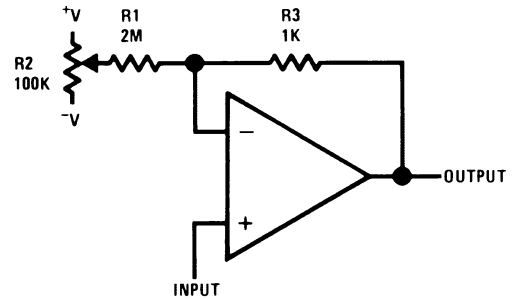
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$$\text{RANGE} = \pm V \left(\frac{R_2}{R_1} \right)$$

$$\text{GAIN} = 1 + \frac{R_5}{R_4 + R_2}$$

FIGURE 3. Offset Voltage Adjustment for Non-Inverting Amplifiers

A voltage follower may be balanced by the technique shown in *Figure 4*. R_1 injects a current which produces a voltage drop across R_3 to cancel the offset voltage. The addition of the adjustment resistors causes a gain error, increasing the gain by 0.05%. This small error usually causes no problem. The adjustment circuit essentially causes the offset voltage to appear at full output, rather than at low output levels, where it is a large percentage error.

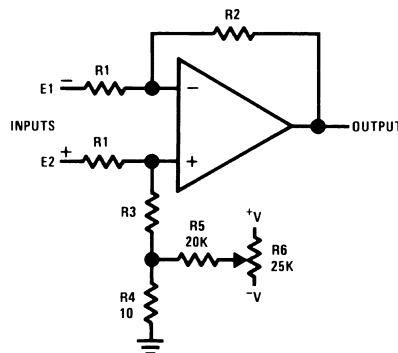


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$$\text{RANGE} = \pm V \left(\frac{R_3}{R_1} \right)$$

FIGURE 4. Offset Voltage Adjustment for Voltage Followers

Differential amplifiers are somewhat more difficult to balance. The offset adjustment used for a differential amplifier can degrade the common mode rejection ratio. *Figure 5* shows an adjustment circuit which has minimal effect on the common mode rejection. The voltage at the arm of the pot is divided by R_4 and R_5 to supply an offset correction of ± 7.5 mV. R_4 and R_5 are chosen such that the common mode rejection ratio is limited by the amplifier for values of R_3 greater than 1 k Ω . If R_3 is less than 1k the shunting of R_4 by R_5 must be considered when choosing the value of R_3 .



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$$R_2 = R_3 + R_4$$

$$\text{RANGE} = \pm V \left(\frac{R_5}{R_4} \right) \left(\frac{R_1}{R_1 + R_3} \right)$$

$$\text{GAIN} = \frac{R_2}{R_1}$$

FIGURE 5. Offset Voltage Adjustment for Differential Amplifiers

The techniques described for balancing offset voltage at the input of the amplifier offer two main advantages: First, they are universally applicable to all operational amplifiers and allow device interchangeability with no modifications to the balance circuitry. Second, they permit balancing without in-

terfering with the internal circuitry of the amplifier. The electrical parameters of the amplifiers are tested and guaranteed without balancing. Although it doesn't usually happen, balancing could degrade performance.

Notes

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